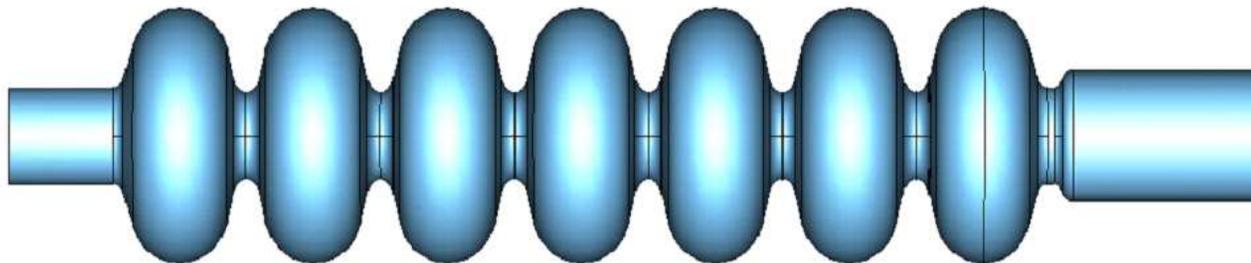


Accelerating (S)RF Cavities

Matthias Liepe





- **Accelerating cavities**
 - DC accelerators
 - RF cavities
 - Examples
- **RF Cavity Fundamentals**
 - Pill box cavity
 - Figures of merit
 - Accelerating mode
 - Multicell cavities and circuit models
- **RF Cavity Design**
 - NC vs. SC
 - Objectives and cell shapes
 - RF codes



- **Accelerating cavities**
 - DC accelerators
 - RF cavities
 - Examples



DC Accelerators

- Use high DC voltage to accelerate particles

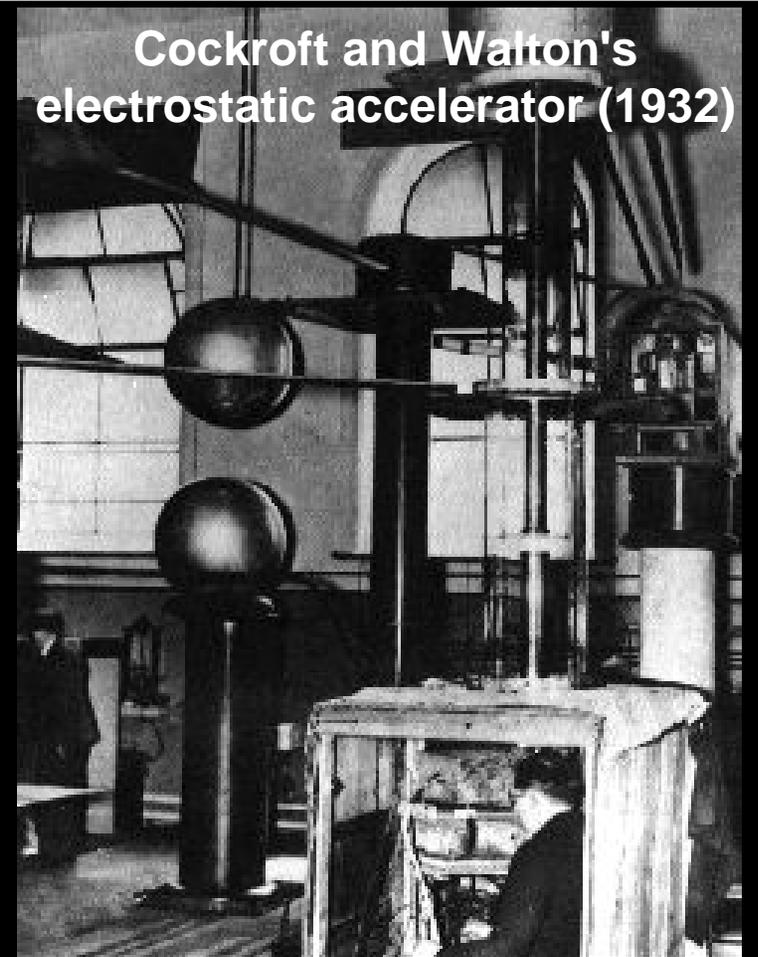
$$\frac{d\vec{p}}{dt} = q(\vec{E} + \vec{v} \times \vec{B})$$

- No work done by magnetic fields

DC accelerator



Cockroft and Walton's electrostatic accelerator (1932)



Protons were accelerated and slammed into lithium atoms producing helium and energy.



DC Accelerators: Limitations

1) DC ($\frac{\partial}{\partial t} \equiv 0$): $\nabla \times \vec{E} = 0$ which is solved by $\vec{E} = -\nabla\Phi$

Limit: If you want to gain 1 MeV, you need a potential of 1 MV!

2) Circular machine: DC acceleration impossible since $\oint \vec{E} \cdot d\vec{s} = 0$

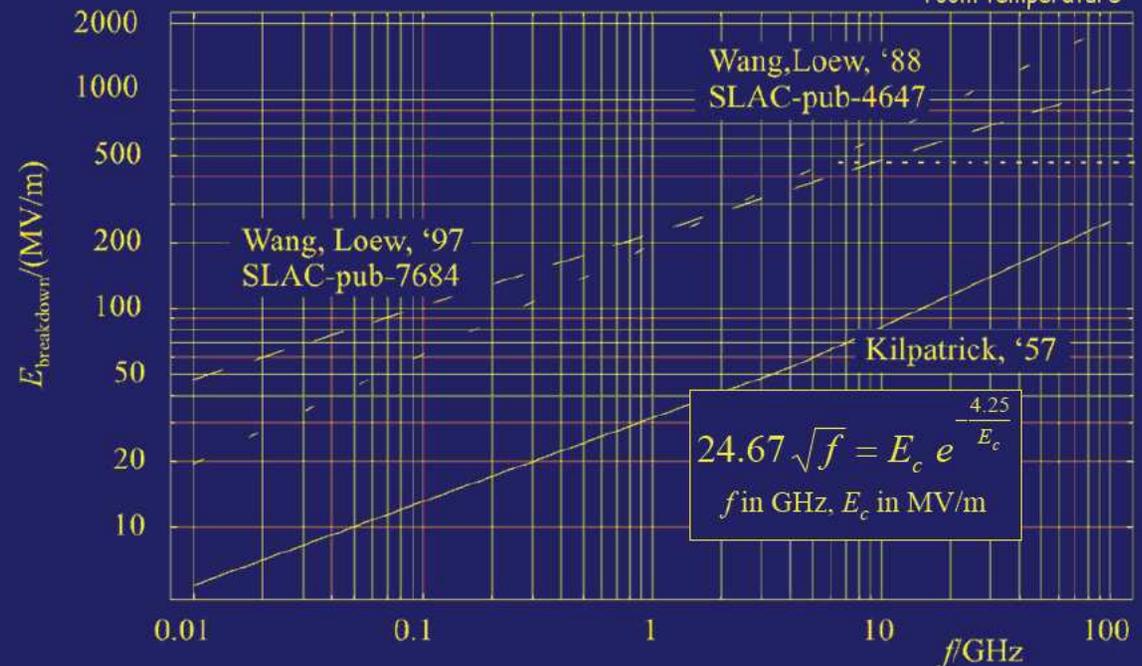
⇒ Use time-varying fields!

Maxwell's equation in vacuum (contd.)

$$\nabla \times \vec{B} - \frac{1}{c^2} \frac{\partial}{\partial t} \vec{E} = 0 \quad \nabla \cdot \vec{B} = 0$$

$$\nabla \times \vec{E} + \frac{\partial}{\partial t} \vec{B} = 0 \quad \nabla \cdot \vec{E} = 0$$

Another reason for RF: breakdown limit in vacuum, Cu surface, room temperature

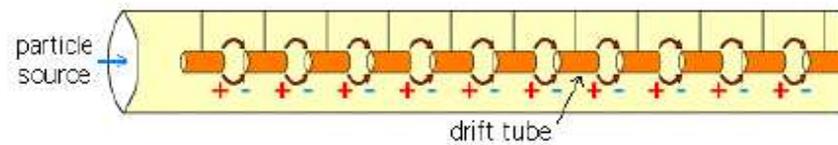




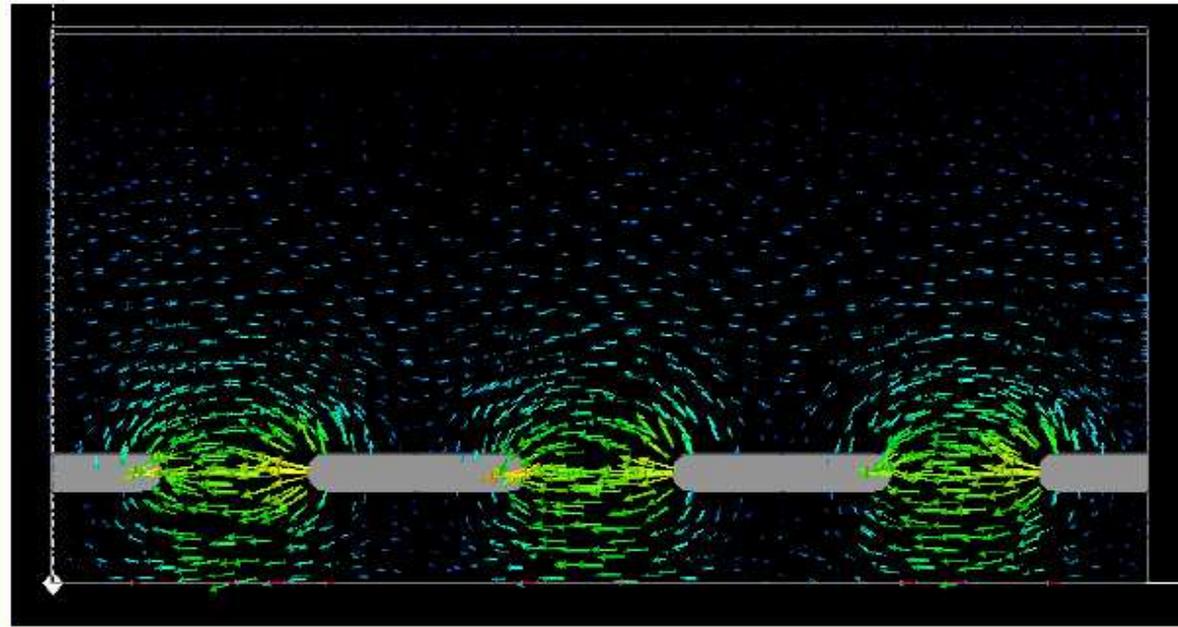
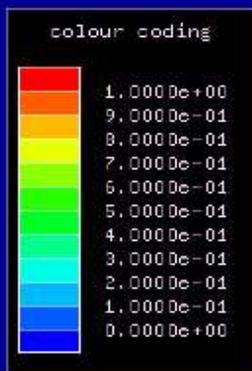
Example of Time Varying Fields: Drift Tube Linac

Drift Tube Linac (DTL) - how it works

For slow particles -
protons @ few MeV e.g.
- the drift tube lengths
can easily be adapted.

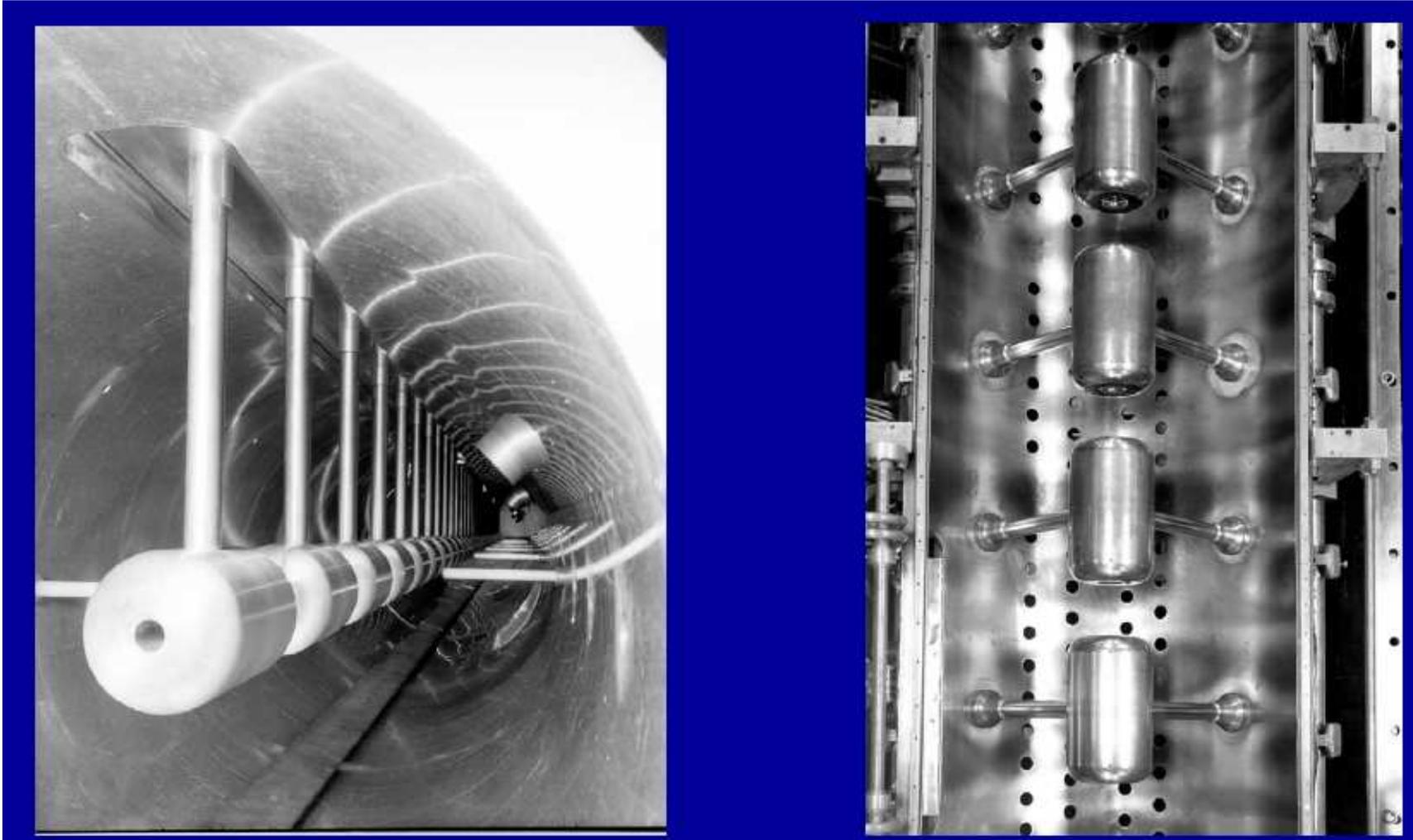


electric field





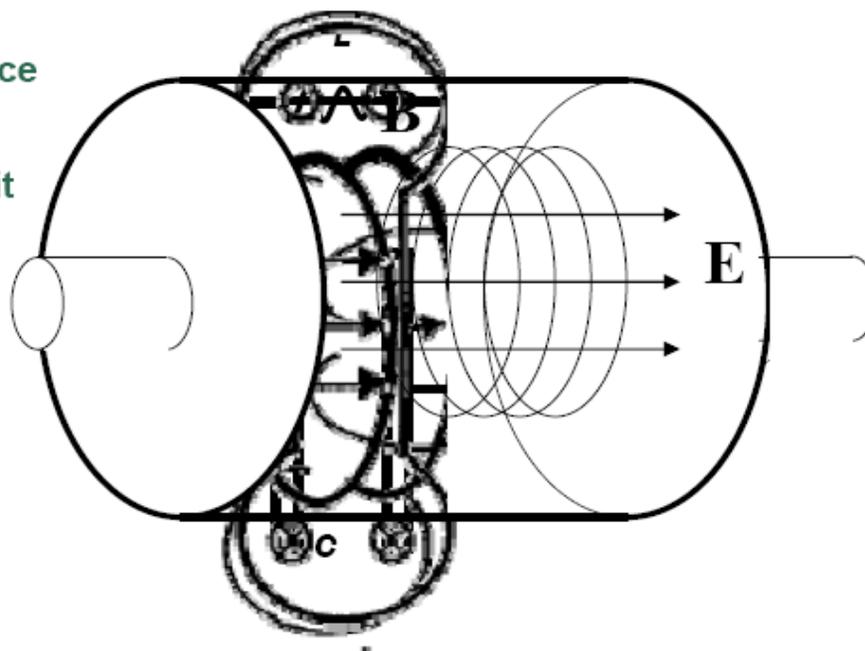
Example of Time Varying Fields: Drift Tube Linac





- For acceleration we require an oscillating RF field
- Simplest form is an LC circuit
- Let $L = 0.1 \text{ mH}$, $C = 0.01 \text{ }\mu\text{F} \rightarrow f = 160 \text{ kHz}$
- To increase the frequency, lower L , eventually only have a single wire
- To reach even lower values must add inductances in parallel
- Eventually have we have a solid wall
- Shorten „wires“ even further to reduce inductance
- \rightarrow Pillbox cavity, „simplest form“
- Add beam tubes to let the particles enter and exit
- Magnetic field concentrated in the cavity wall, losses will be here.

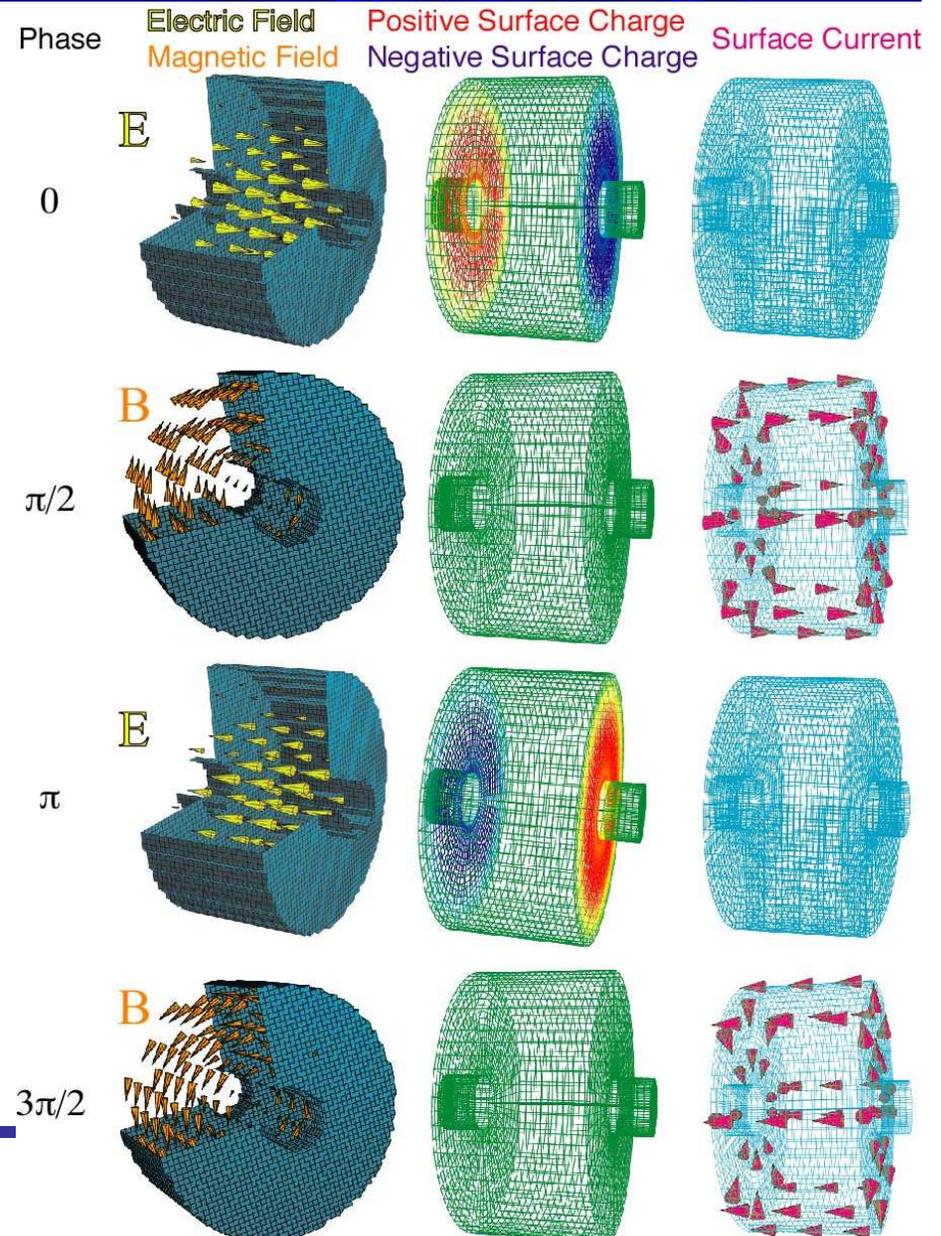
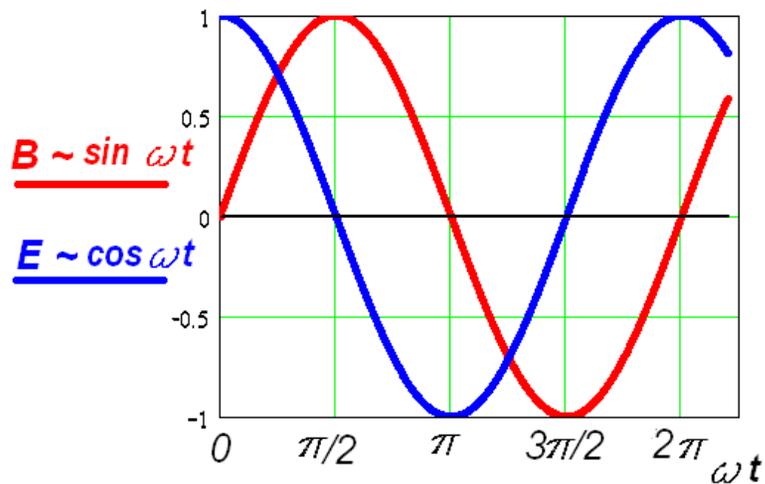
Based on Feynman's Lect. on Physics.





RF Cavities

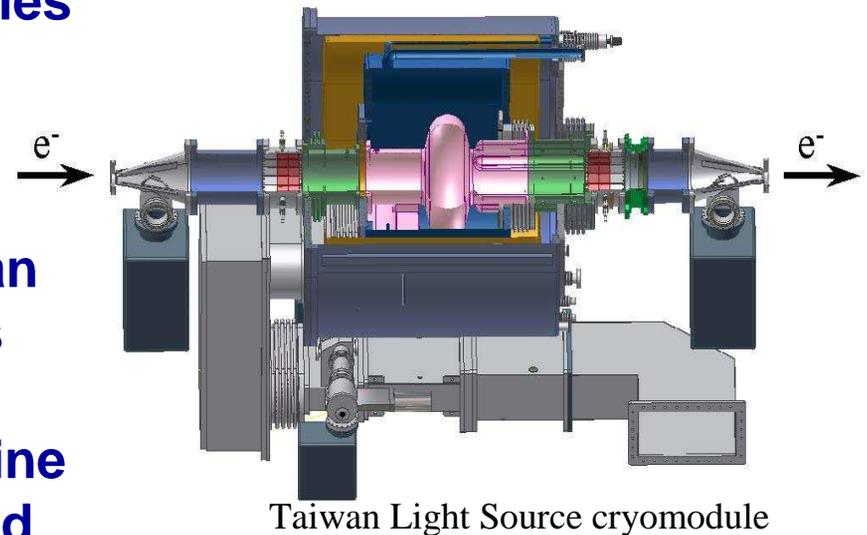
- Time dependent electromagnetic field inside metal box





Introduction to RF Cavities

- The main purpose of using RF cavities in accelerators is to provide energy gain to charged-particle beams
- The highest achievable gradient, however, is not always optimal for an accelerator. There are other factors (both machine-dependent and technology-dependent) that determine operating gradient of RF cavities and influence the cavity design, such as accelerator cost optimization, maximum power through an input coupler, necessity to extract HOM power, etc.
- In many cases requirements are competing.



Taiwan Light Source cryomodule



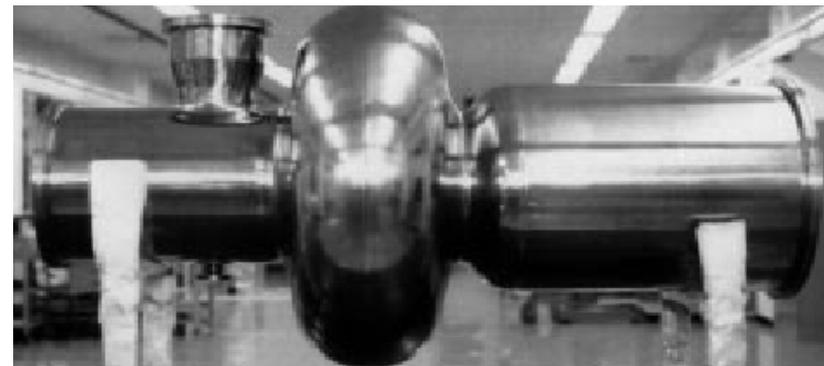
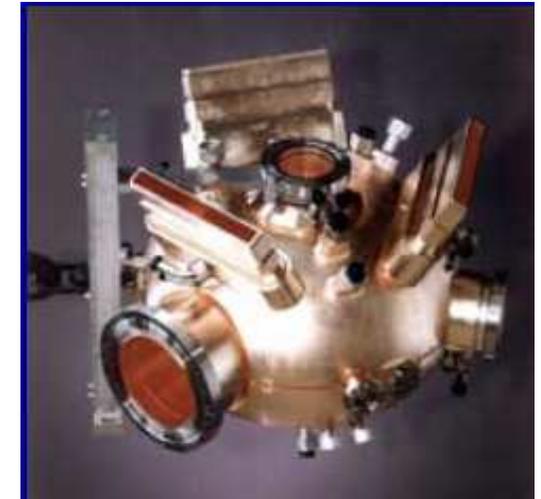
CW High-Current Storage Rings (colliders and light sources)

- NC or SC
- Relatively low gradient
(1...9 MV/m)
- Strong HOM damping
($Q \sim 10^2$)
- High average RF power
(hundreds of kW)



CESR cavities

PEP II Cavity

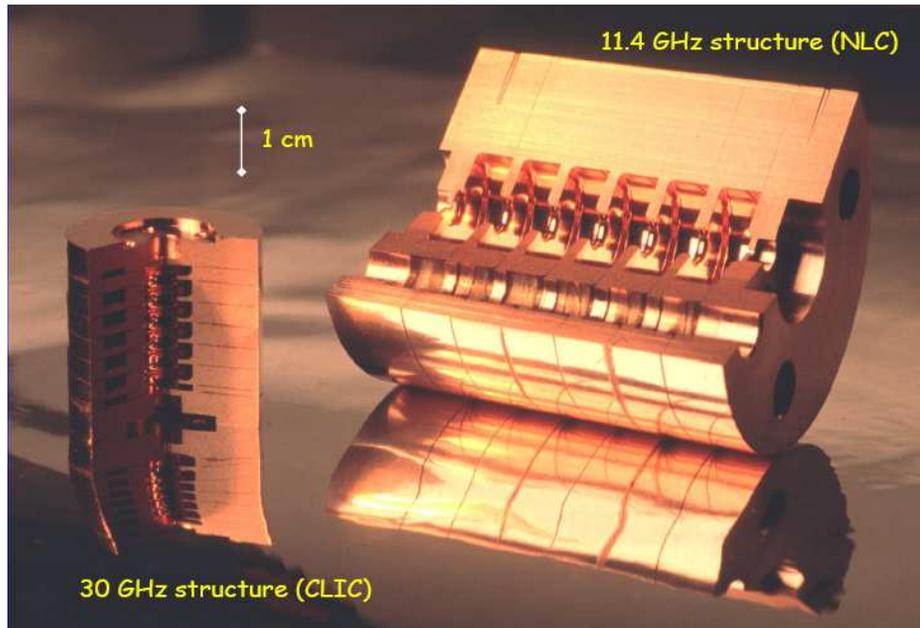
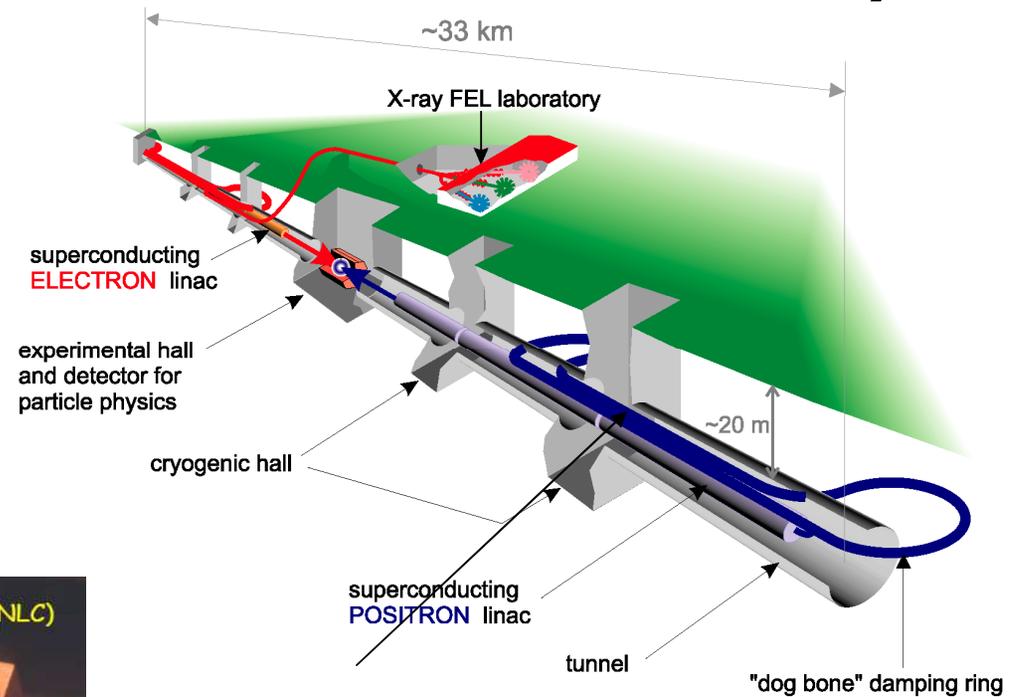


KEK cavity



Pulsed Linacs (ILC, XFEL, ...)

- High gradients
- Moderate HOM damping reqs.
- High peak RF power

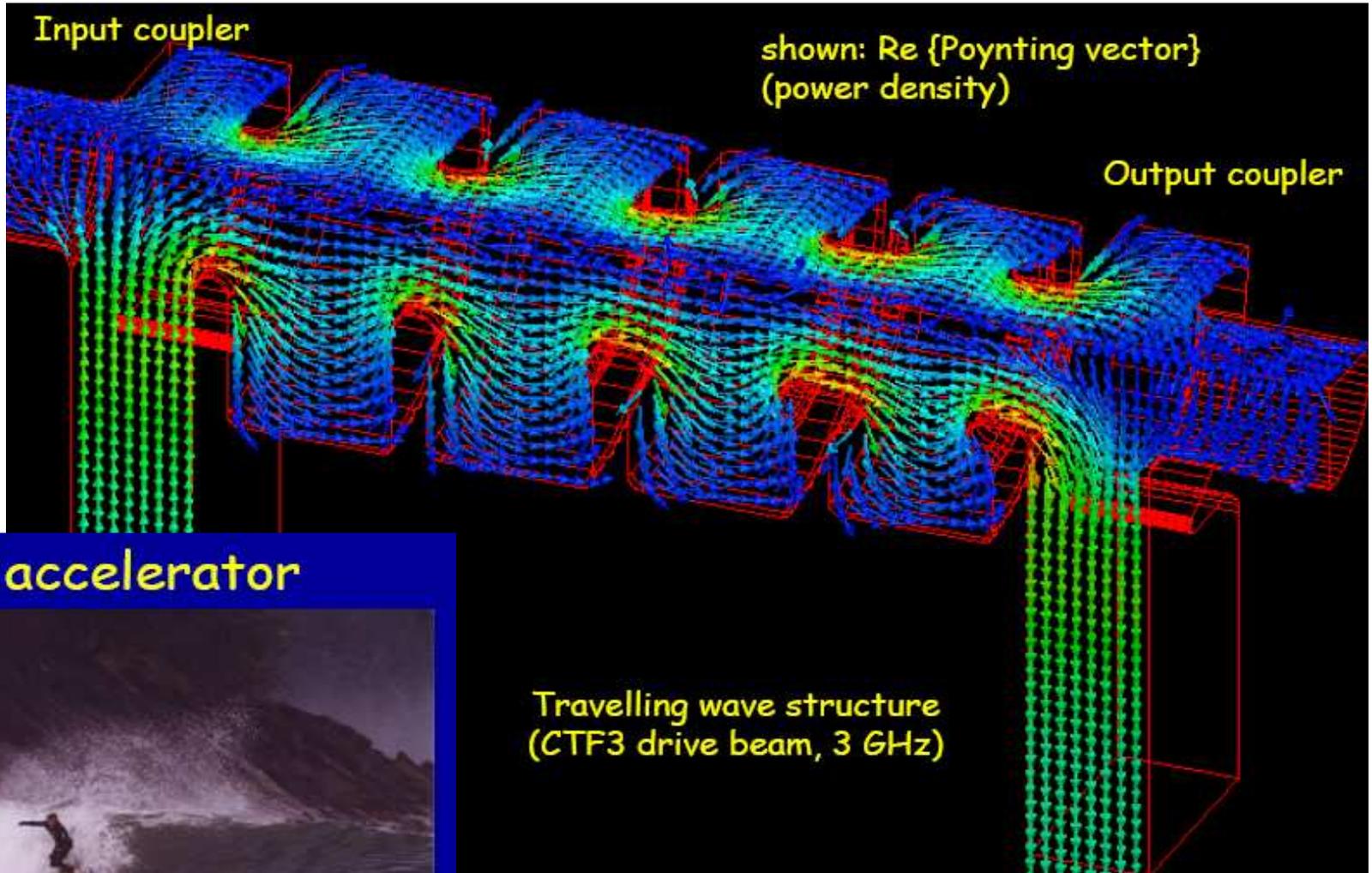


ILC: 21,000 cavities!
ILC / XFEL cavities





Traveling Wave Cavities



RF accelerator

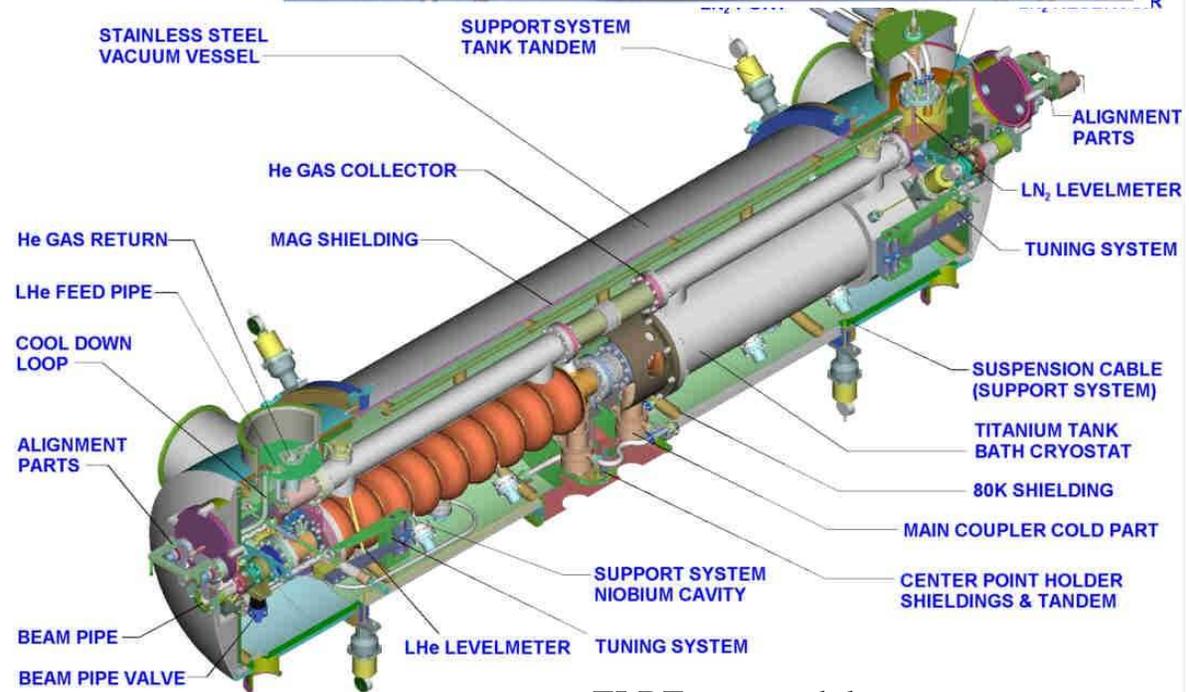
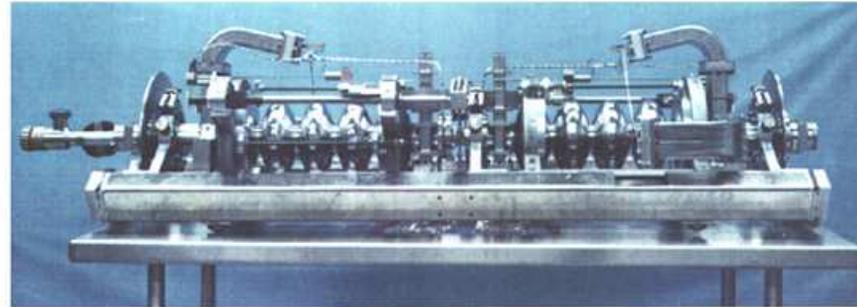




CW low-current linacs (CEBAF, ELBE)

- SRF cavities
- Moderate to low gradient (8...20 MV/m)
- Relaxed HOM damping requirements
- Low average RF power (5...13 kW)

CEBAF cavities

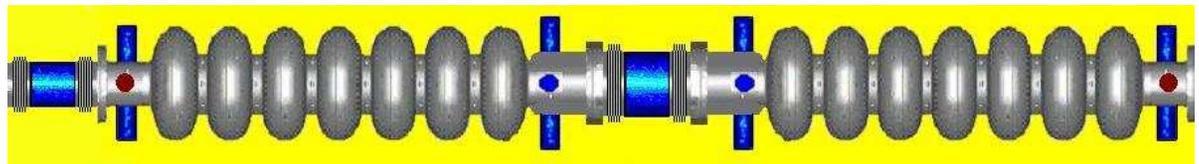
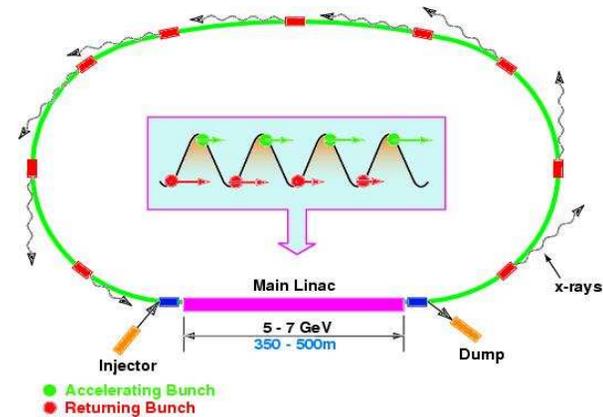


ELBE cryomodule

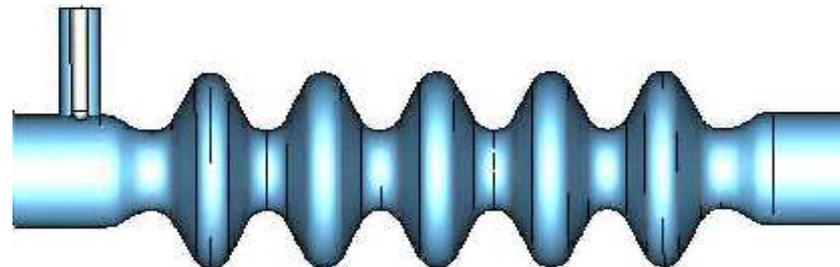


CW High-Current ERLs

- SRF cavities
- Moderate gradient (15...20 MV/m)
- Strong HOM damping ($Q = 10^2 \dots 10^4$)
- Low average RF power (few kW)



Cornell ERL cavities



BNL ERL cavity



- **RF Cavity Fundamentals (Standing wave cavities)**
 - Cavity eigenmodes
 - Figures of merit
 - Accelerating mode
 - Multicell cavities and circuit model



RF Cavities and their Eigenmodes I

Cavity \equiv an arbitrary volume, partially closed by the metal wall, capable to store the E-H energy



~ 3.95 GHz is the lowest frequency

First assumption:

1. Stored E-H fields are harmonic in time.

Maxwell equations for the harmonic, lossless case with no free charge in the volume

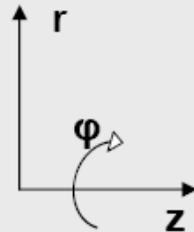
$$\left\{ \begin{array}{l} \nabla \times H = i\omega\epsilon E \\ \nabla \times E = -i\omega\mu H \\ \nabla \cdot E = 0 \\ \nabla \cdot H = 0 \end{array} \right.$$



RF Cavities and their Eigenmodes II

Second assumption (good approximation for the elliptical cavities):

2. The volume is cylindrically symmetric. We commonly use the (r, φ, z) coordinates.



z is conventional direction of the acceleration and symmetry axis

$$\begin{cases} \nabla_c \times H = i\omega\epsilon E \\ \nabla_c \times E = -i\omega\mu H \\ \nabla_c \cdot E = 0 \\ \nabla_c \cdot H = 0 \end{cases}$$

$$\nabla_c \times A = \vec{i}_r \left(\frac{1}{r} \frac{\partial A_z}{\partial \varphi} - \frac{\partial A_\varphi}{\partial z} \right) + \vec{i}_\varphi \left(\frac{\partial A_r}{\partial z} - \frac{\partial A_z}{\partial r} \right) + \vec{i}_z \left(\frac{1}{r} \frac{\partial (rA_\varphi)}{\partial r} - \frac{1}{r} \frac{\partial A_r}{\partial \varphi} \right)$$

$$\nabla_c \cdot A = \frac{1}{r} \frac{\partial (rA_r)}{\partial r} + \frac{1}{r} \frac{\partial A_\varphi}{\partial \varphi} + \frac{\partial A_z}{\partial z}$$



RF Cavities and their Eigenmodes III

3. For the acceleration are suitable field patterns with strong E along the beam trajectory. This ensures, by the proper phasing, maximal energy exchange between the cavity and beam.

TM_{0xx}-like monopole modes have “very strong” E_z component on the symmetry axis.

Fields of the monopole modes are independent on φ .

$$\frac{\partial E}{\partial \varphi} = 0 \quad \frac{\partial H}{\partial \varphi} = 0$$

Non monopole (HOM) modes have component $E_z = 0$ on the symmetry axis.

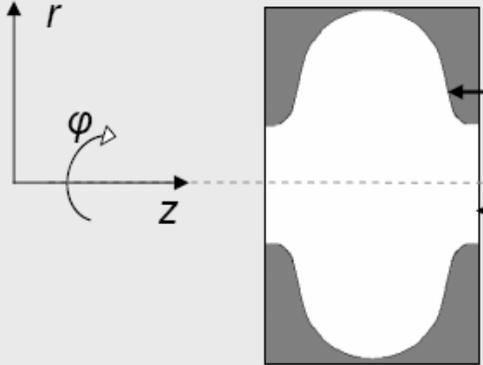
Their fields dependent on φ .



RF Cavities and their Eigenmodes IV

Maxwell equations + boundary conditions for E and H fields lead to the Helmholtz equation, which is an eigenvalue problem.

For H(r,z) field of a monopole mode the equation is:


$$(\nabla_c^2 + \omega^2 \epsilon \mu) H = 0$$
$$n \cdot H = 0 \quad \text{on metal wall}$$
$$\begin{cases} H = 0 \\ n \cdot H = 0 \end{cases} \quad \text{optionally on non metal boundary}$$
$$\nabla_c^2 A = \nabla_c (\nabla_c \cdot A) - \nabla_c \times \nabla_c \times A$$

There is infinity number of TM_{0xx} solutions (modes) to the Helmholtz equation.

All modes are determine by:

$$H_n(r,z) = [0, H_{\phi,n}(r,z), 0],$$

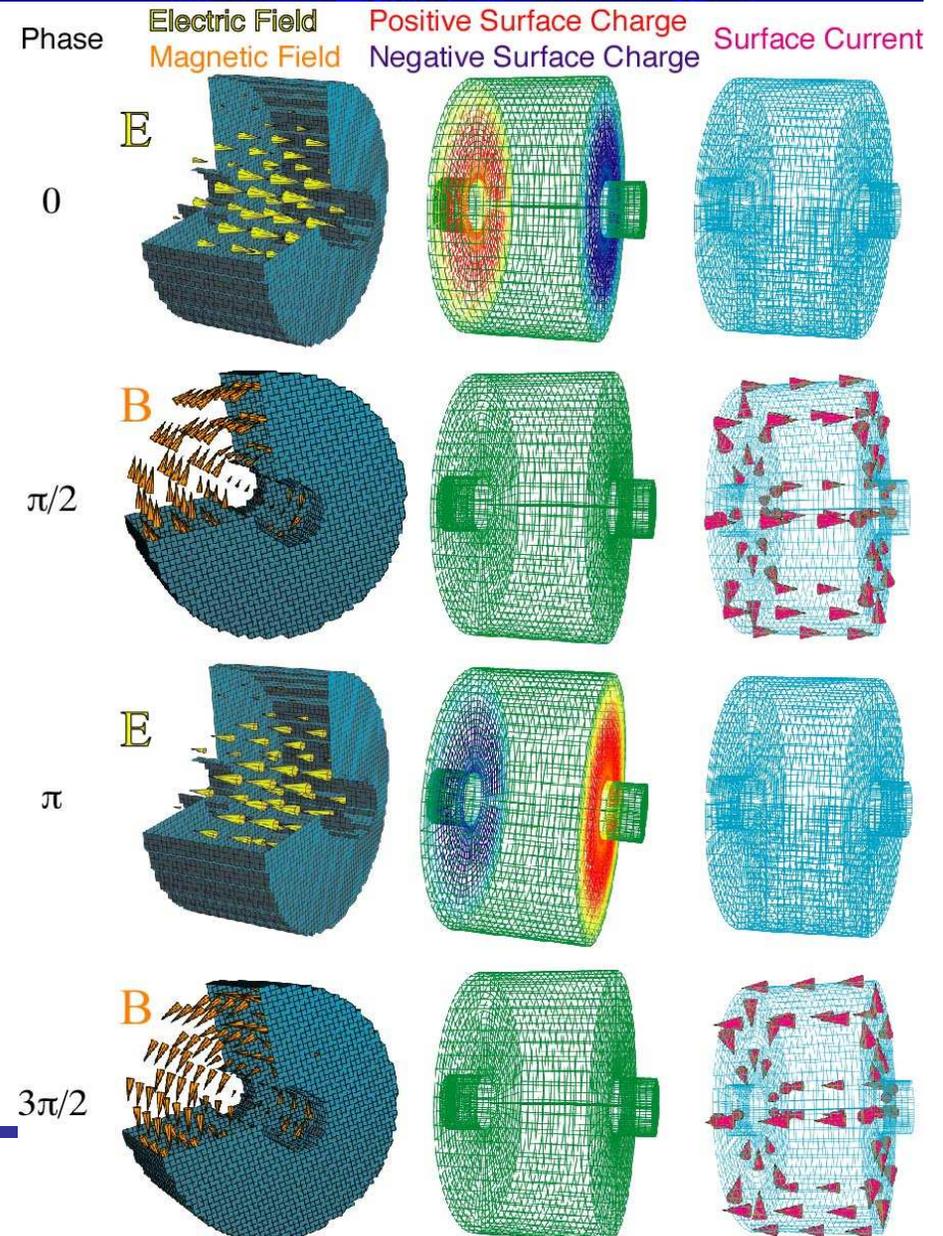
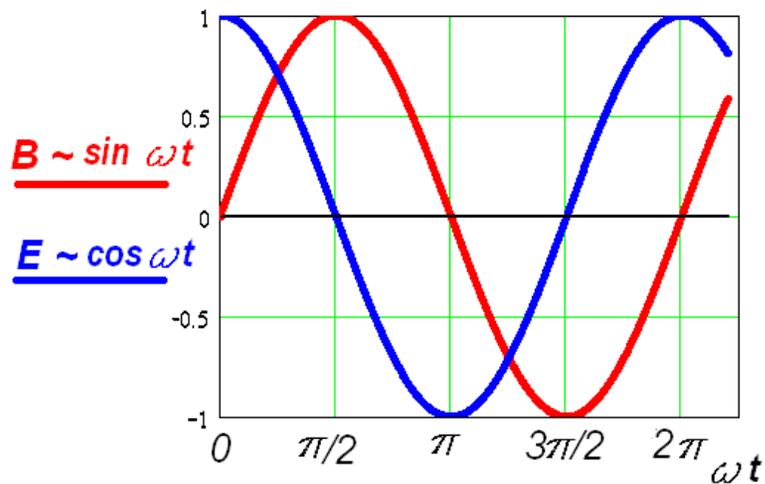
$$E_n(r,z) = [E_{r,n}(r,z), 0, E_{z,n}(r,z)]$$

and frequency ω_n .



Accelerating Mode

- Time dependent electromagnetic field inside metal box





Figures of Merit - 1

Accelerating Voltage & Accelerating Field ($v = c$ for Particles)

- Frequency
- For maximum acceleration we need

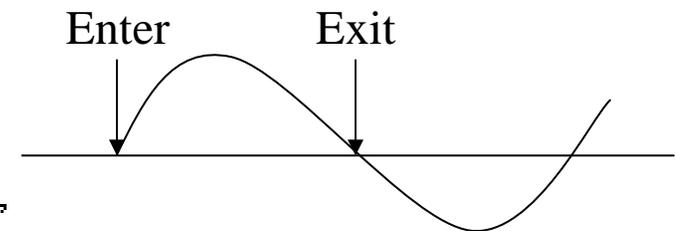
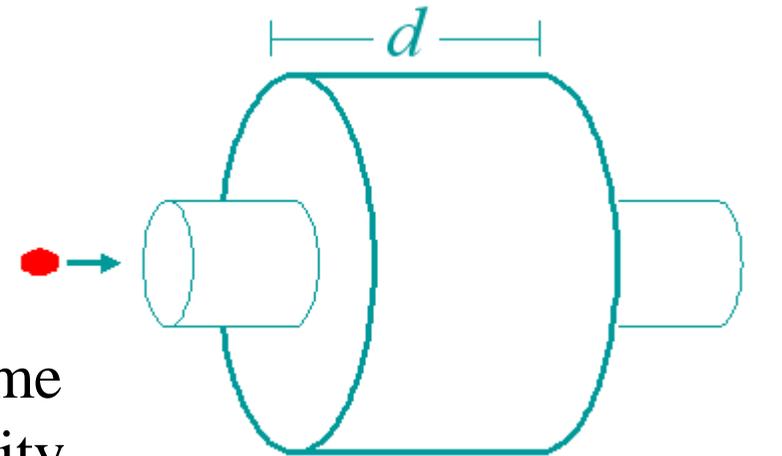
$$T_{\text{cav}} = \frac{d}{c} = \frac{T_{\text{RF}}}{2}$$

so that the field always points in the same direction as the bunch traverses the cavity

- Accelerating voltage then is

$$V_{\text{cav}} = \text{Re} \left[\int_0^d E_z(\rho=0, z) e^{i\omega_0 z/c} dz \right] = d \cdot E_0 \frac{\sin \frac{\omega_0 d}{c}}{\frac{\omega_0 d}{c}} = d \cdot E_0 T$$

- Accelerating field is $E_{\text{acc}} = \frac{V_{\text{cav}}}{d} = 2E_0/\pi$



, here, $T = 2/\pi$
 is flight time factor
 (for pill-box with this length d only, other shapes can have a different value of T).



- Surface currents ($\propto H$) result in dissipation proportional to the surface resistance (R_s):

$$\frac{dP_c}{ds} = \frac{1}{2} R_s |\mathbf{H}|^2$$

- Dissipation in the cavity wall given by surface integral:

$$P_c = \frac{1}{2} R_s \int_S |\mathbf{H}|^2 ds$$

- Stored energy is: $\longrightarrow U = \frac{1}{2} \mu_0 \int_V |\mathbf{H}|^2 dv$

- Define Quality (Q) as $Q_0 = \frac{\omega_0 U}{P_c} = 2 \pi \frac{U}{T_{\text{rf}} P_c}$

which is $\sim 2 \pi$ number of cycles it takes to dissipate the energy stored in the cavity \rightarrow Easy way to measure Q

- $Q_{nc} \approx 10^4$, $Q_{sc} \approx 10^{10}$



Figures of Merit for Cavity Design - 3 Geometry Factor - G

Since the time averaged energy in the electric field equals that in magnetic field, the total energy in the cavity is given by

$$U = \frac{1}{2}\mu_0 \int_V |\mathbf{H}|^2 dv = \frac{1}{2}\epsilon_0 \int_V |\mathbf{E}|^2 dv,$$

where the integral is taken over the volume of the cavity.

Power dissipated in the cavity walls is $P_c = \frac{1}{2}R_s \int_S |\mathbf{H}|^2 ds$,

where the integration is taken over the interior cavity surface.

$$Q_0 = \frac{\omega_0 U}{P_c}, \quad Q_0 = \frac{\omega_0 \mu_0 \int_V |\mathbf{H}|^2 dv}{R_s \int_S |\mathbf{H}|^2 ds}.$$

The Q_0 is frequently written as $Q_0 = \frac{G}{R_s}$,

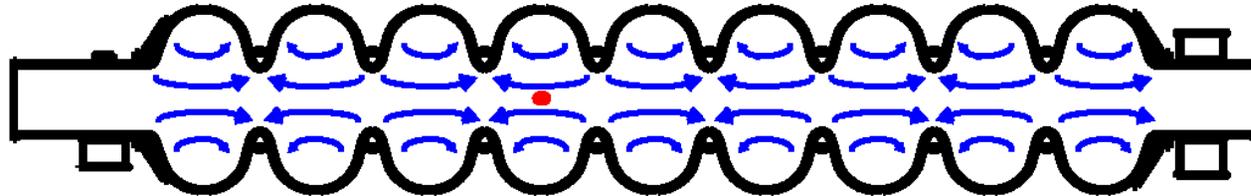
where $G = \frac{\omega_0 \mu_0 \int_V |\mathbf{H}|^2 dv}{\int_S |\mathbf{H}|^2 ds}$. G is known as the geometry

constant. From the last eq. we can see that it depends on the cavity shape but not its size.



Figures of Merit for Cavity Design - 4 Shunt Impedance (R_a)

Accelerating π -mode:



Accelerating voltage:

$$V_{acc} = \frac{\text{maximum energy gain}}{\text{charge}} = \int_{-L/2}^{+L/2} E_z e^{i\omega(z/c)} dz$$

Shunt-impedance:

$$R_a = \frac{(V_{acc})^2}{P_{dis}}$$

Quality factor:

$$Q_L = \frac{\omega U}{P_{dis}}$$

$$\frac{R_a}{Q_L} = \frac{R}{Q} = \frac{(V_{acc})^2}{\omega U}$$



Figures of Merit for Cavity Design - 4 Shunt Impedance (R_a)

- Shunt impedance (R_a) determines how much acceleration one gets for a given dissipation (analogous to Ohm's Law)

$$R_a = \frac{V_c^2}{P_c}$$

→ To maximize acceleration (P_c given), must maximize shunt impedance.

Another important figure of merit is $\frac{R_a}{Q_0} = \frac{V_c^2}{\omega_0 U}$,

- R_a/Q only depends on the cavity geometry

→ This quantity is also used for determining the level of mode excitation by charges passing through the cavity.

→ To minimize losses (P_c) in the cavity, we must maximize $G \cdot R_a/Q_0$:

$$P_c = \frac{V_c^2}{R_a} = \frac{V_c^2}{Q_0 \cdot (R_a/Q_0)} = \frac{V_c^2}{(R_s \cdot Q_0)(R_a/Q_0)/R_s} = \frac{V_c^2 \cdot R_s}{G \cdot (R_a/Q_0)}$$

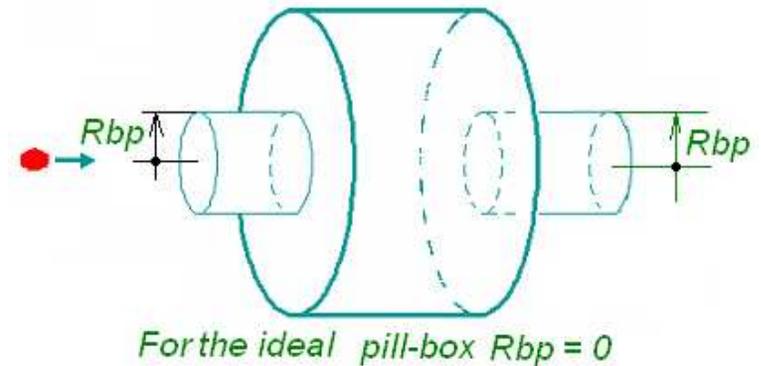
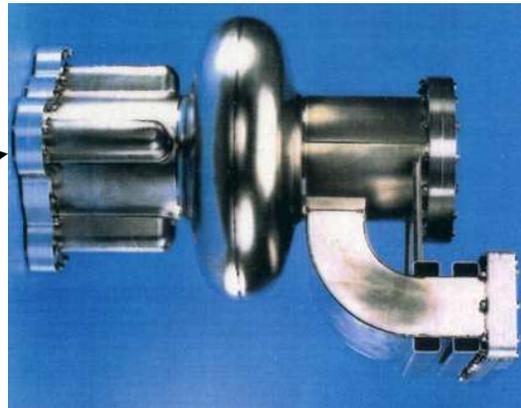


Typical Values for Single Cells

Quantity	Cornell SC 500 MHz	Pillbox
G	270 Ω	257 Ω
R_a/Q_0	88 Ω /cell	196 Ω /cell
E_{pk}/E_{acc}	2.5	1.6
H_{pk}/E_{acc}	52 Oe/MV/m	30.5 Oe/(MV/m)

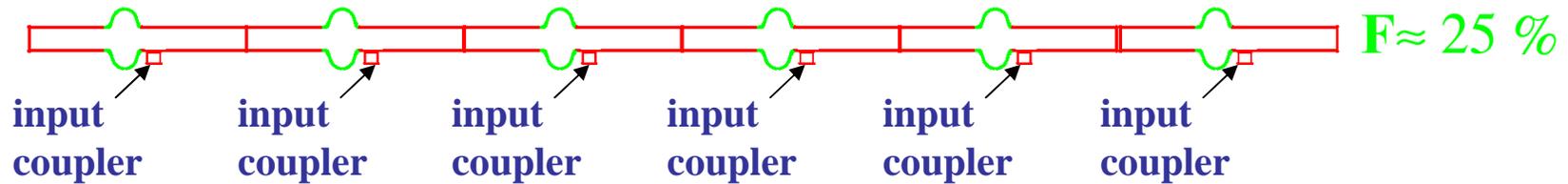
Current is high, it excites a lot of Higher Order Modes, so the hole is made big to propagate HOMs, and this is why H_{pk} and E_{pk} grew, and R/Q drops.

Diff. applications – diff. trade-offs.





Multicell Cavities: Why?



higher fill-factor:

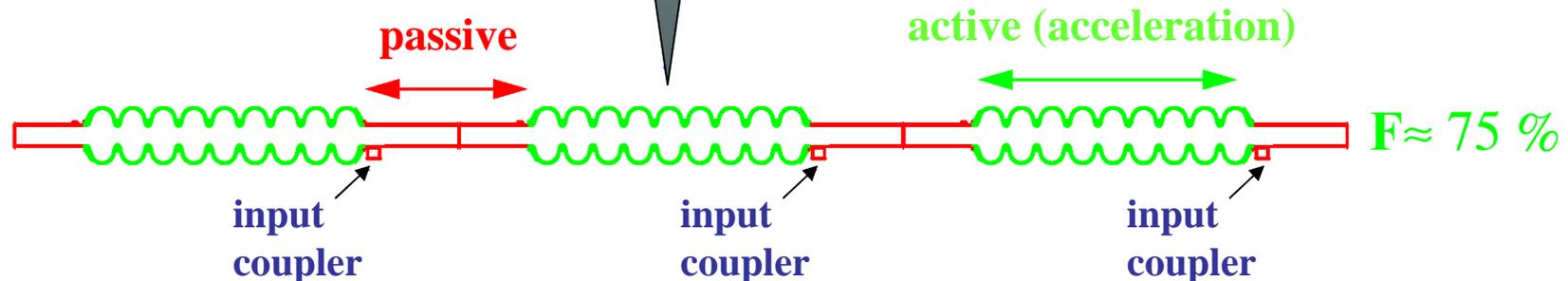
$$F = \frac{\text{active length}}{\text{total length}}$$

\Rightarrow *lower costs*

\Rightarrow *better beam*

fewer

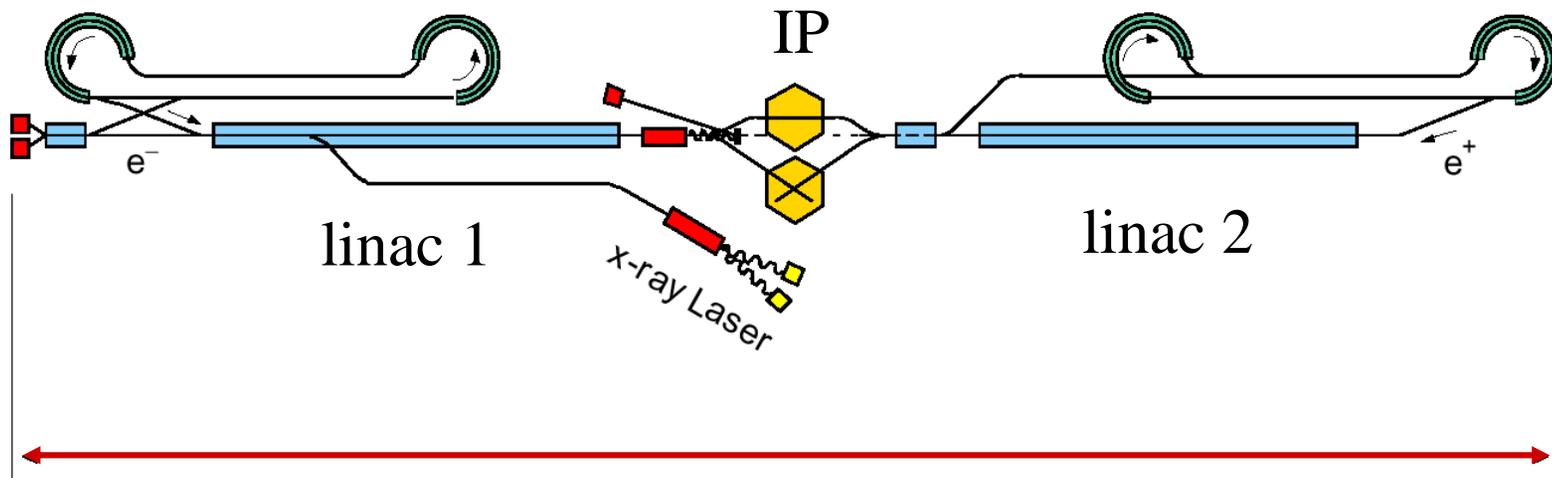
- input couplers
- waveguide elements
- RF control systems
- ...





Multicell Cavities: Why?

Example: 500 GeV Linear Collider



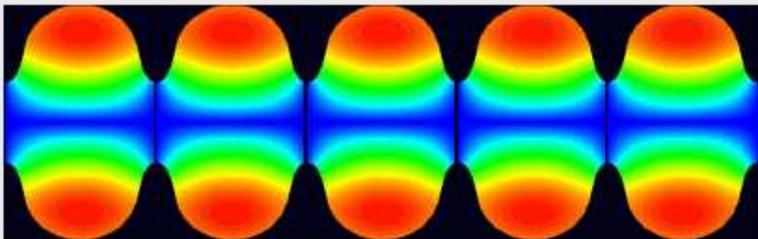
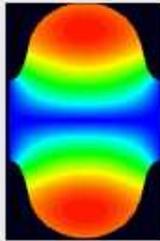
21,024 9-cell cavities: 27.8 km (17.3 miles)

189,216 1-cell cavities: 75.4 km (46.8 miles)



Multicell Cavities and Cell-to-Cell Coupling

The last parameter, relevant for multi-cell accelerating structures, is the coupling k_{cc} between cells for the accelerating mode passband (Fundamental Mode passband).



Single-cell structures are attractive from the RF-point of view:

- ➔ Easier to manage HOM damping
- ➔ No field flatness problem.
- ➔ Input coupler transfers less power
- ➔ Easy for cleaning and preparation
- ➔ **But it is expensive to base even a small linear accelerator on the single cell. We do it only for very high beam current machines.**

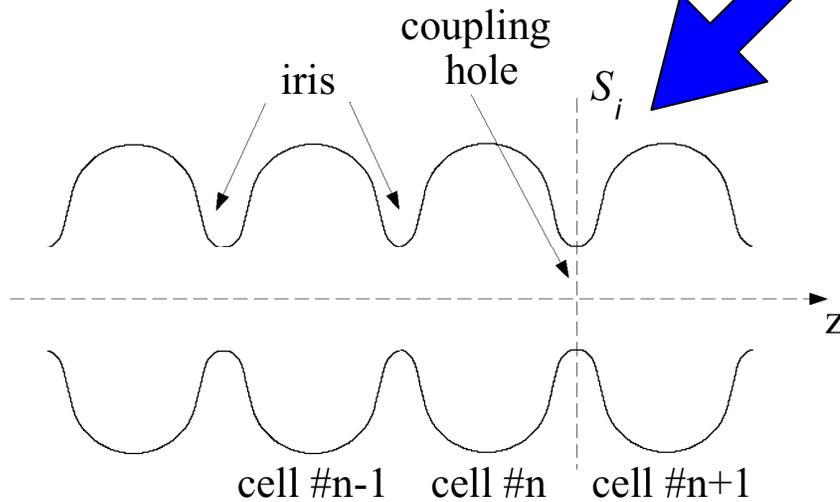
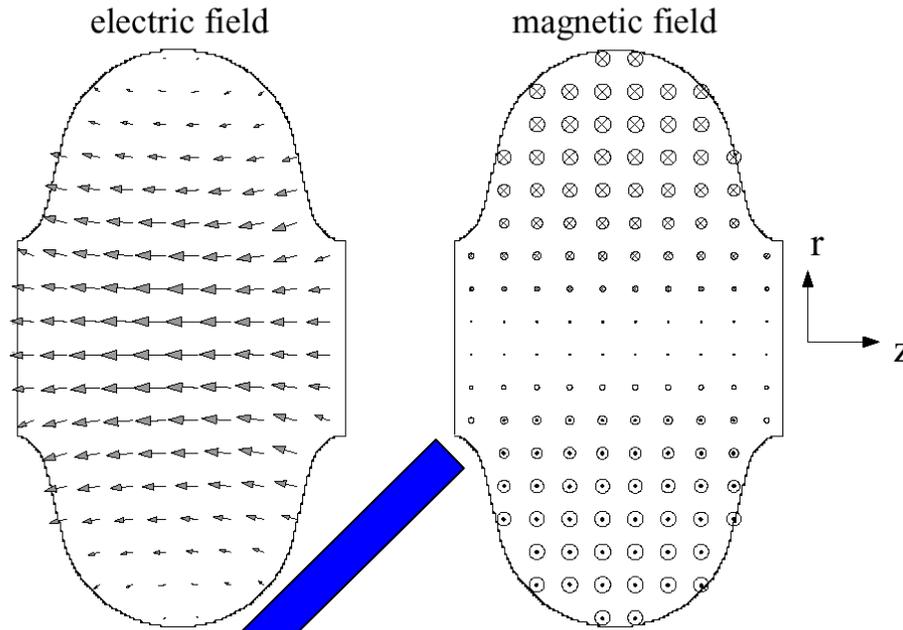
Multi-cell structures are less expensive and offers higher real-estate gradient but:

- ➔ **Field flatness (stored energy) in cells becomes sensitive to frequency errors of individual cells**
- ➔ **Other problems arise: HOM trapping...**



Cell-to-Cell Coupling

Single Cell:
TM₀₁₀ mode

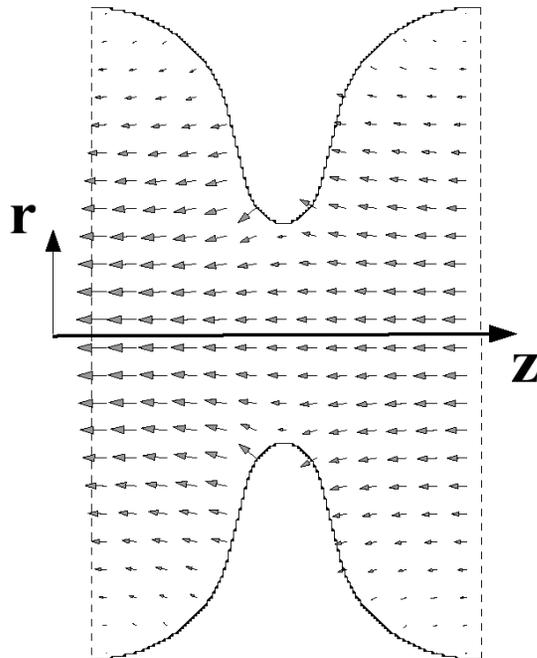


Coupled Cells

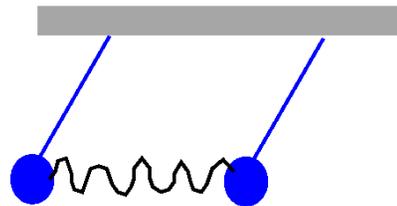
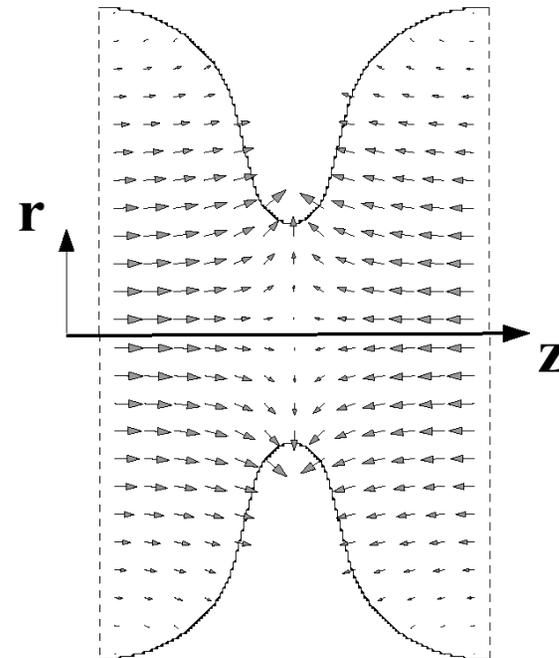


Two Coupled Cells: TM₀₁₀ Modes

0 - Mode

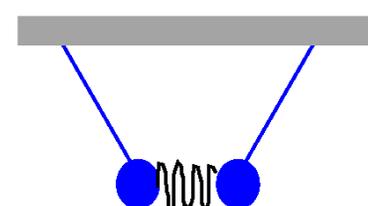


π - Mode



2 coupled cells

n coupled cells

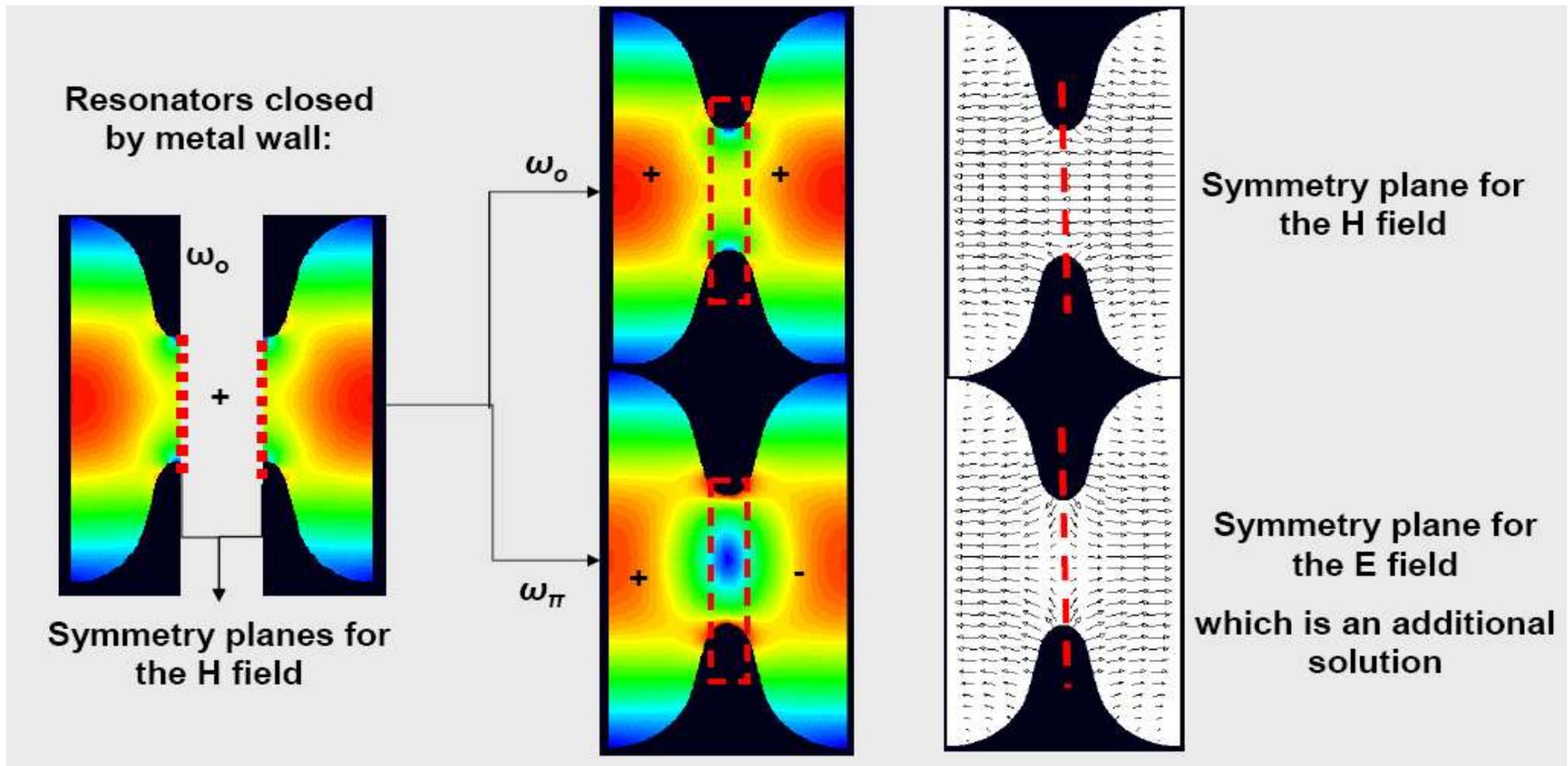


2 TM₀₁₀ modes

n TM₀₁₀ modes

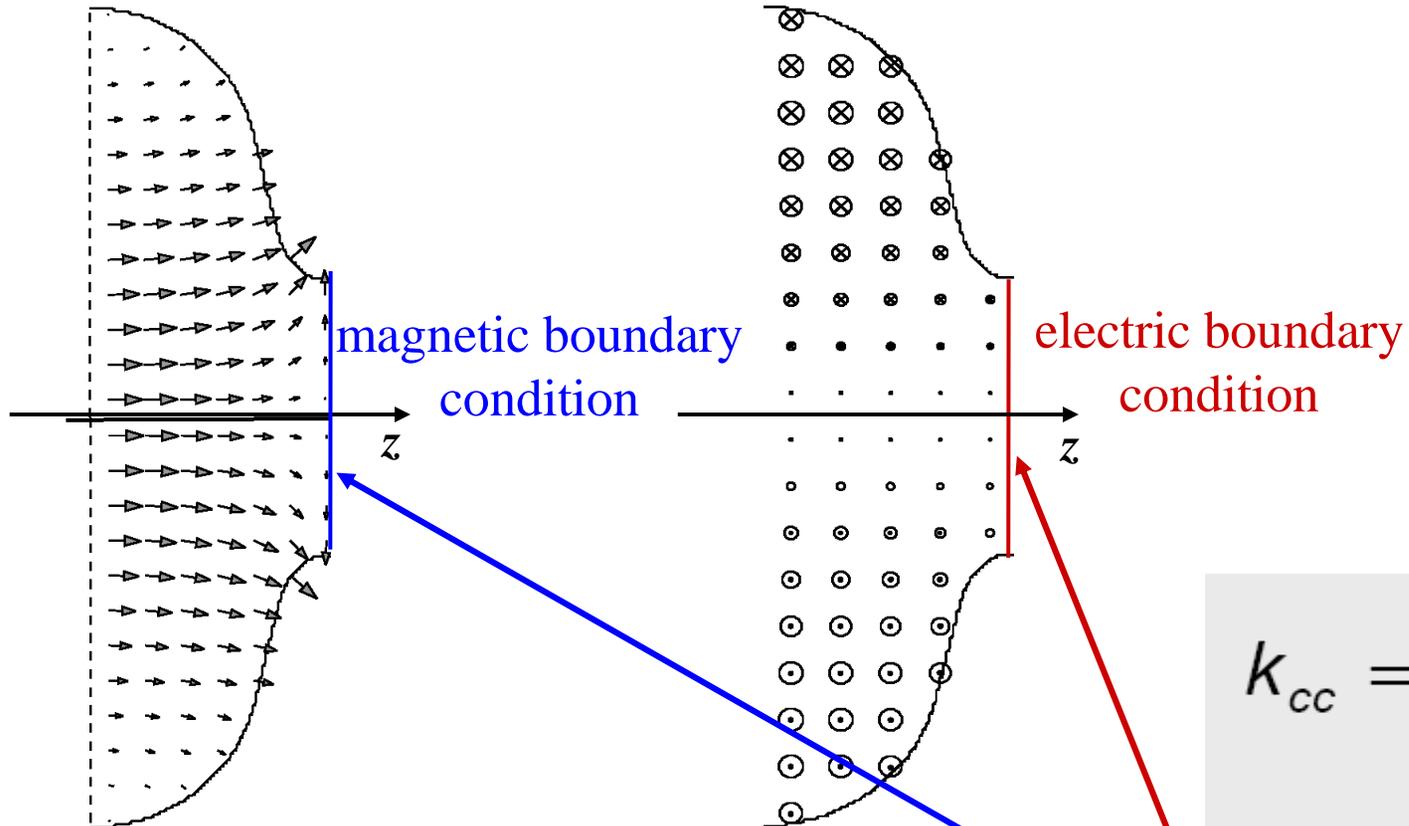


Two Coupled Cells: TM₀₁₀ Modes





Cell-to-Cell Coupling



$$k_{cc} = \frac{\omega_{\pi} - \omega_0}{\frac{\omega_{\pi} + \omega_0}{2}}$$

Coupling factor:

$$k_{cc} = \frac{c}{\omega_0} \int_{S_i} (\vec{e} \times \vec{h}) \cdot \vec{u}_z ds$$



A Circuit Model: Let's start at the Beginning: Step 1.1

Solve **Maxwell's equations** for given boundary conditions



orthogonal eigenmodes

$$\vec{E}^{(m)}(\vec{r}, t) = \vec{E}^{(m)}(\vec{r})e^{i\omega^{(m)}t}$$
$$\vec{H}^{(m)}(\vec{r}, t) = \vec{H}^{(m)}(\vec{r})e^{i\omega^{(m)}t + i\pi/2}, m = 1, 2, \dots$$



orthonormal eigenfunctions

$$\vec{e}^{(m)}(\vec{r}) = \sqrt{\frac{\epsilon_0}{2U^{(m)}}} \vec{E}^{(m)}(\vec{r})$$

$$\int_V \vec{e}^{(m)}(\vec{r}) \vec{e}^{(n)}(\vec{r}) dv = \delta_{mn}$$

$$\vec{h}^{(m)}(\vec{r}) = \sqrt{\frac{\mu_0}{2U^{(m)}}} \vec{H}^{(m)}(\vec{r})$$

$$\int_V \vec{h}^{(m)}(\vec{r}) \vec{h}^{(n)}(\vec{r}) dv = \delta_{mn}$$



A Circuit Model: Step 1.2

With Maxwell's equations in vacuum:

$$\vec{\nabla} \times \vec{H} = \varepsilon_0 \frac{\partial \vec{E}}{\partial t} \quad \vec{\nabla} \times \vec{E} = -\mu_0 \frac{\partial \vec{H}}{\partial t}$$

and the eigenmodes

$$\vec{E}^{(m)}(\vec{r}, t) = \vec{E}^{(m)}(\vec{r}) e^{i\omega^{(m)}t}$$

$$\vec{H}^{(m)}(\vec{r}, t) = \vec{H}^{(m)}(\vec{r}) e^{i\omega^{(m)}t + i\pi/2}$$

and the normalized eigenfunctions

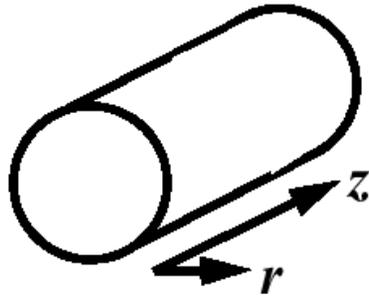
$$\vec{h}^{(m)}(\vec{r}) = \sqrt{\frac{\mu_0}{2U^{(m)}}} \vec{H}^{(m)}(\vec{r}) \quad \vec{e}^{(m)}(\vec{r}) = \sqrt{\frac{\varepsilon_0}{2U^{(m)}}} \vec{E}^{(m)}(\vec{r})$$

we get the relations

$$\vec{\nabla} \times \vec{h}^{(m)} = \frac{\omega^{(m)}}{c} \vec{e}^{(m)} \quad \vec{\nabla} \times \vec{e}^{(m)} = \frac{\omega^{(m)}}{c} \vec{h}^{(m)}$$



A Circuit Model: Step 1 Eigenmodes: Example: Pillbox-Cavity



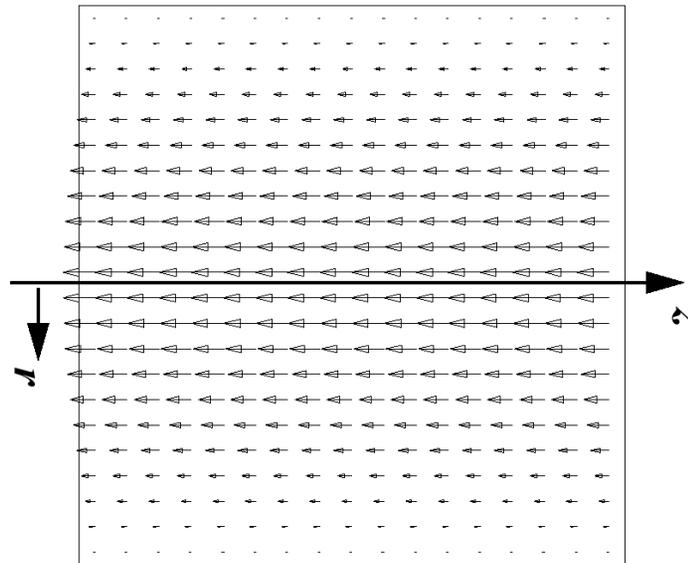
lowest frequency mode:

transverse magnetic

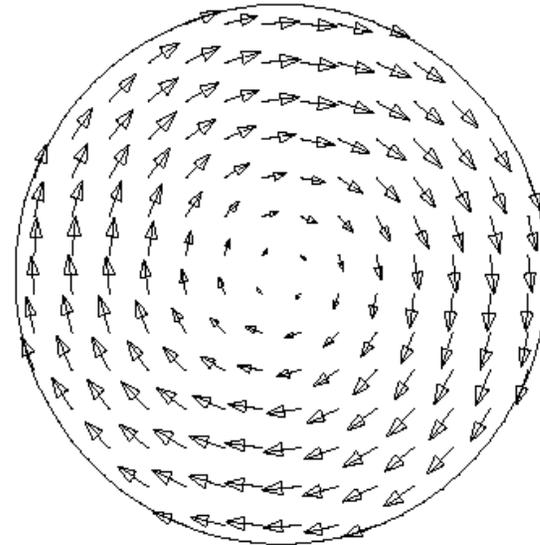
TM₀₁₀

0 \Rightarrow monopole mode

electric field



magnetic field



$$\vec{E}(r, t) = E_0 J_0\left(\frac{\omega_0}{c} r\right) \cos(\omega_0 t) \vec{e}_z \quad \Rightarrow \text{acceleration}$$

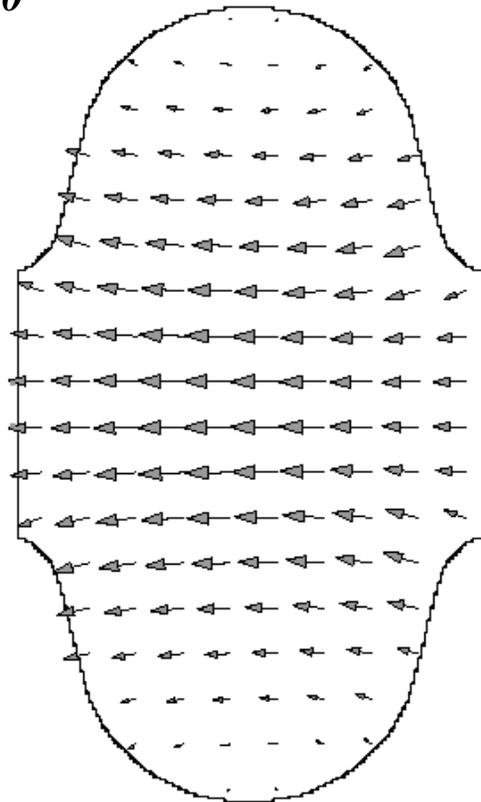


A Circuit Model: Step 1

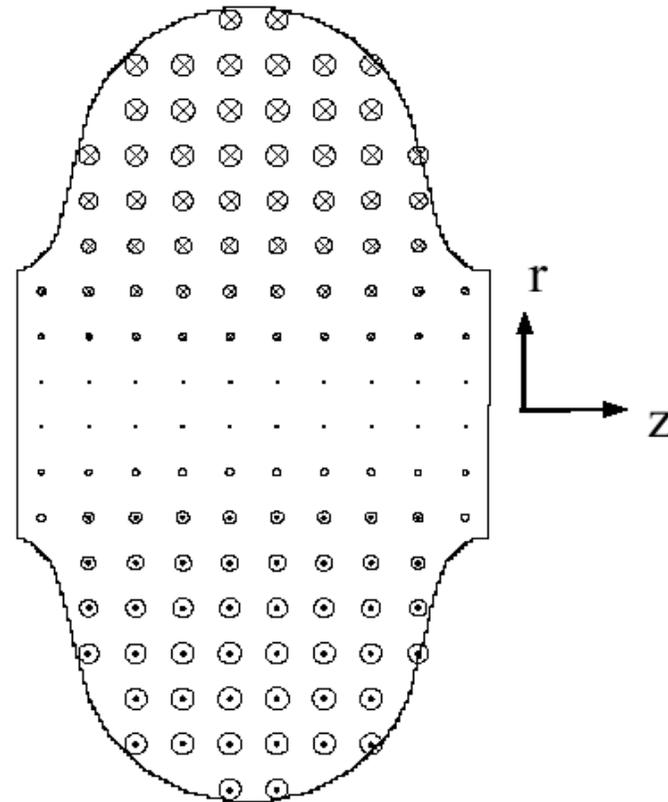
Eigenmodes: Example: Single Cell TESLA Cavity

Accelerating mode:

TM_{010} : *electric field*



magnetic field





Circuit Model Step 2: Expansion in Eigenmodes

Each *time dependent field* in a cavity can be written as
a *sum of the cavity eigenfunctions*
with *time dependent amplitudes*:

$$\vec{E}(\vec{r}, t) = \sum_m \hat{E}^{(m)}(t) \vec{e}^{(m)}(\vec{r})$$

*time dependent
amplitude*

eigenfunction

$$\vec{H}(\vec{r}, t) = \sum_m \hat{H}^{(m)}(t) \vec{h}^{(m)}(\vec{r})$$

*time dependent
amplitude*

eigenfunction



Insert expansion of fields into Maxwell's equations in vacuum:

$$\vec{\nabla} \times \vec{H} = \epsilon_0 \frac{\partial \vec{E}}{\partial t}$$

insert eigenmode-
expansion of fields...



...and the equations
from page 37

$$\sum_m \hat{H}^{(m)}(t) \frac{\omega^{(m)}}{c} \vec{e}^{(m)}(\vec{r}) = \sum_m \epsilon_0 \frac{d\hat{E}^{(m)}(t)}{dt} \vec{e}^{(m)}(\vec{r})$$

use orthogonality of
eigenmodes:



multiply by $\vec{e}^{(m)}$ and
integrate over cavity
volume

$$\frac{\omega^{(m)}}{c} \hat{H}^{(m)}(t) - \epsilon_0 \frac{d}{dt} \hat{E}^{(m)}(t) = 0 \quad , m = 1, 2, \dots$$



Insert expansion of fields into Maxwell's equations:

$$\vec{\nabla} \times \vec{E} = -\mu_0 \frac{\partial \vec{H}}{\partial t}$$

insert eigenmode-
expansion of fields...



...and the equations
from page 37

$$\sum_m \hat{E}^{(m)}(t) \frac{\omega^{(m)}}{c} \vec{h}^{(m)}(\vec{r}) = -\sum_m \mu_0 \frac{d\hat{H}^{(m)}(t)}{dt} \vec{h}^{(m)}(\vec{r})$$

use orthogonality of
eigenmodes:



multiply by $\vec{h}^{(m)}$ and
integrate over cavity
volume

$$\frac{\omega^{(m)}}{c} \hat{E}^{(m)}(t) + \mu_0 \frac{d}{dt} \hat{H}^{(m)}(t) = 0 \quad , m = 1, 2, \dots$$



Differential Equation for the Eigenmode Amplitudes

$$\frac{\omega^{(m)}}{c} \hat{H}^{(m)}(t) - \epsilon_0 \frac{d}{dt} \hat{E}^{(m)}(t) = 0 \quad , m = 1, 2, \dots$$

$$\frac{\omega^{(m)}}{c} \hat{E}^{(m)}(t) + \mu_0 \frac{d}{dt} \hat{H}^{(m)}(t) = 0 \quad , m = 1, 2, \dots$$



$$\frac{d^2}{dt^2} \hat{E}^{(m)}(t) + \left(\omega^{(m)}\right)^2 \hat{E}^{(m)}(t) = 0 \quad , m = 1, 2, \dots$$

*amplitude of
eigenmode #m*

⇒ Differential equation of an oscillator (we will add losses and a generator later)!



TM₀₁₀ modes in coupled cells:

N coupled cells ⇒ N coupled oscillators

⇒ N coupled differential equations:

$$\frac{d^2}{dt^2} \hat{E}_n + \omega_n^2 \hat{E}_n + \omega_n^2 \frac{K}{2} (\hat{E}_n - \hat{E}_{n-1}) + \omega_n^2 \frac{K}{2} (\hat{E}_n - \hat{E}_{n+1}) = 0 \quad , n = 1, 2, \dots, N$$

TM₀₁₀
eigenfrequency
of cell #n

cell-to-cell
coupling factor

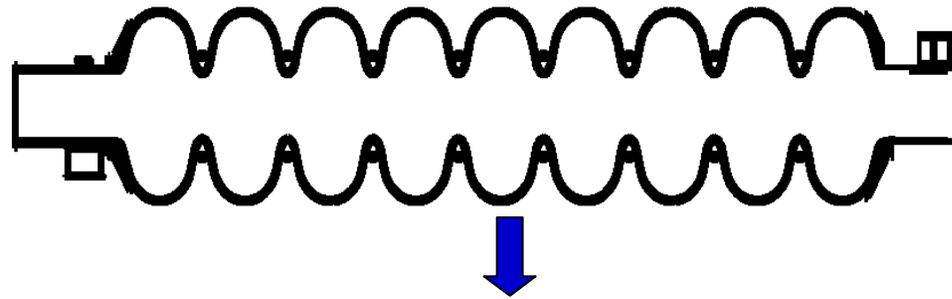
TM₀₁₀ field
amplitude in
cell #n

N-cell structure
⇒ N coupled
differential equations



Circuit Model: Step 5

Equivalent Circuit Model (TM₀₁₀ Modes only)

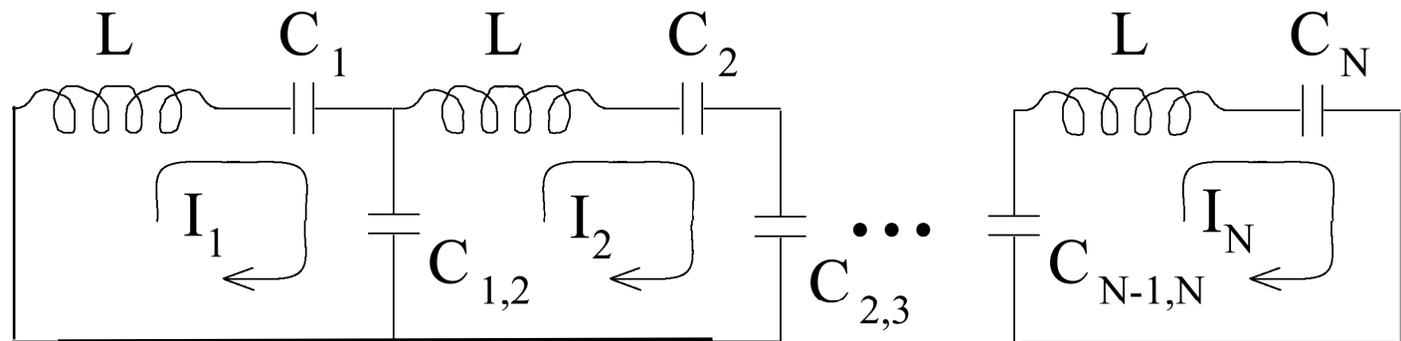


$$\frac{d^2}{dt^2} \hat{E}_n + \omega_n^2 \hat{E}_n + \omega_n^2 \frac{K}{2} (\hat{E}_n - \hat{E}_{n-1}) + \omega_n^2 \frac{K}{2} (\hat{E}_n - \hat{E}_{n+1}) = 0 \quad , n = 1, 2, \dots, N$$

substitute:

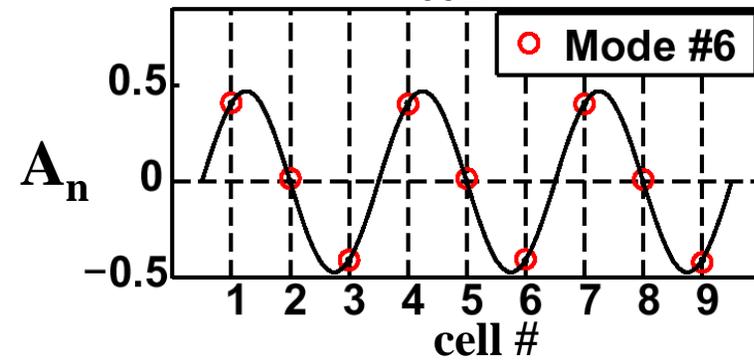
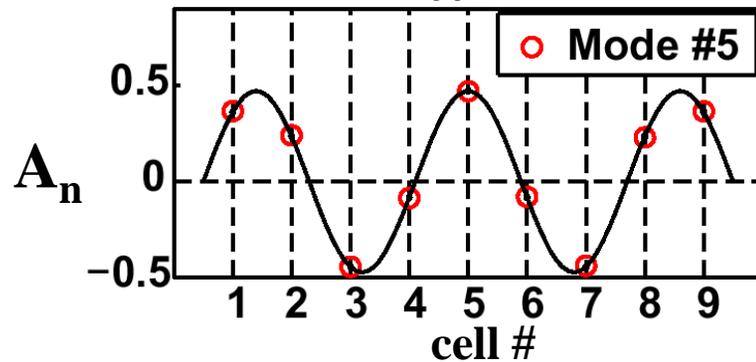
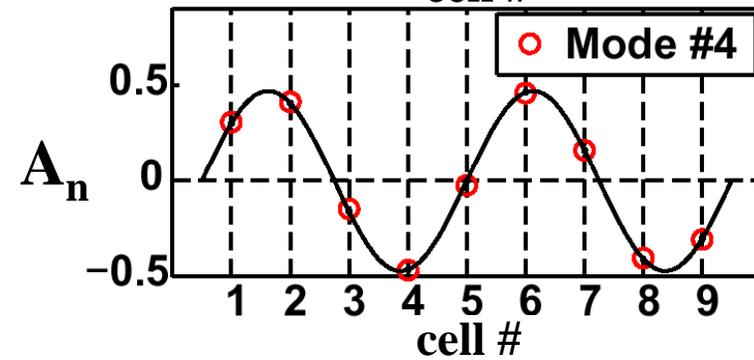
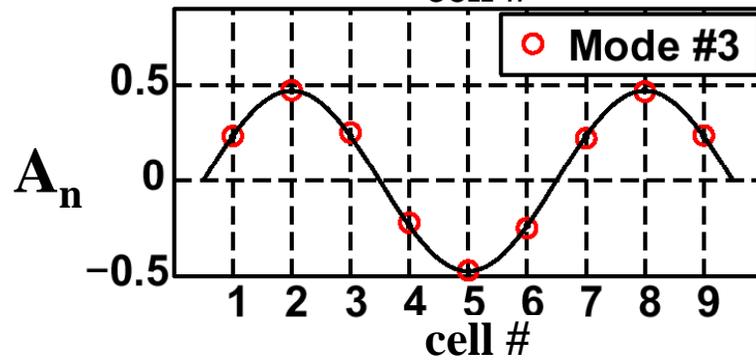
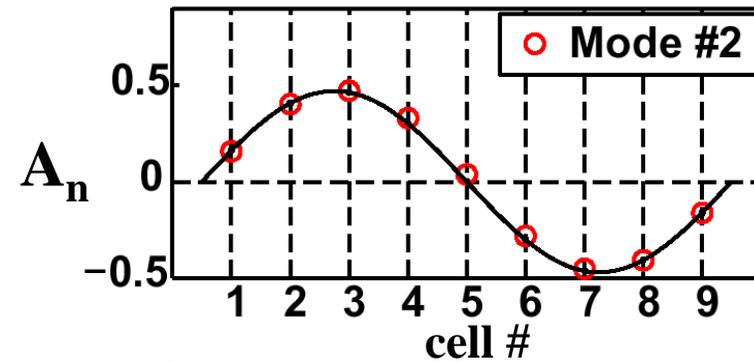
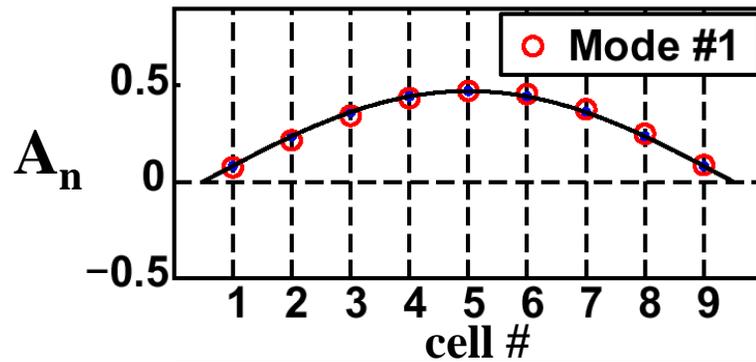
$$\omega_n^2 = \frac{1}{LC_n}$$

$$K_{n,n+1} = 2 \frac{C_n}{C_{n,n+1}} = 2k_{n,n+1}$$



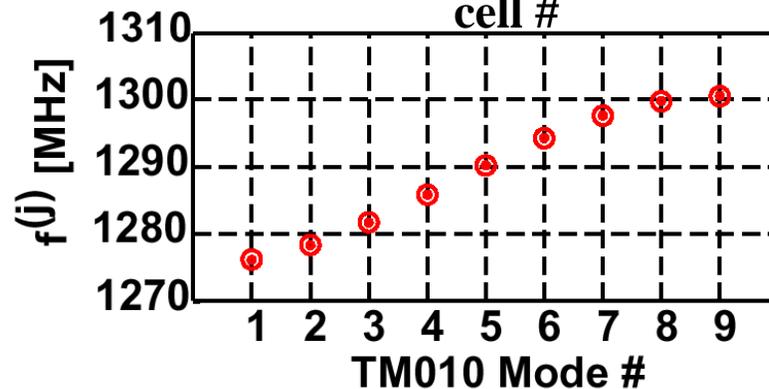
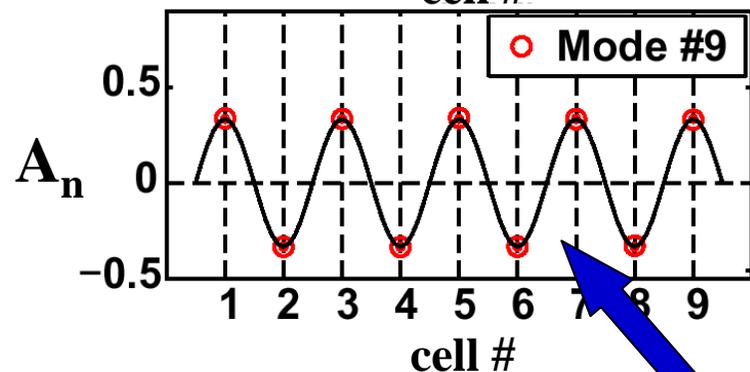
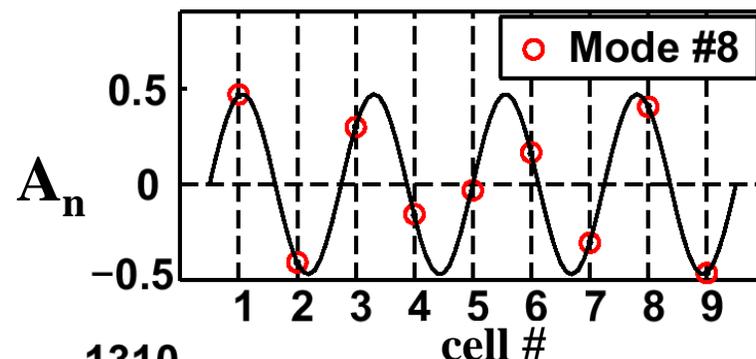
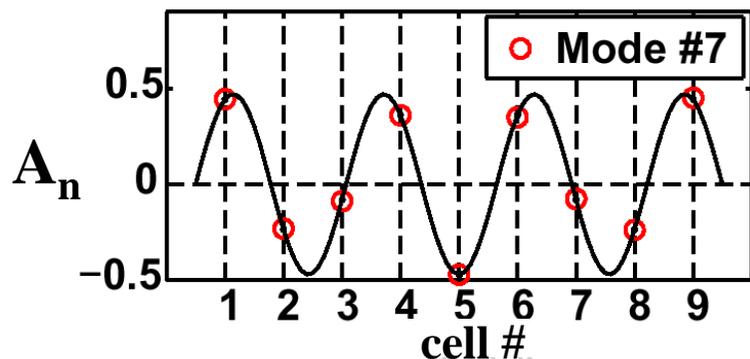


TM₀₁₀ Eigenmodes Example: 9-Cell Cavity (1)





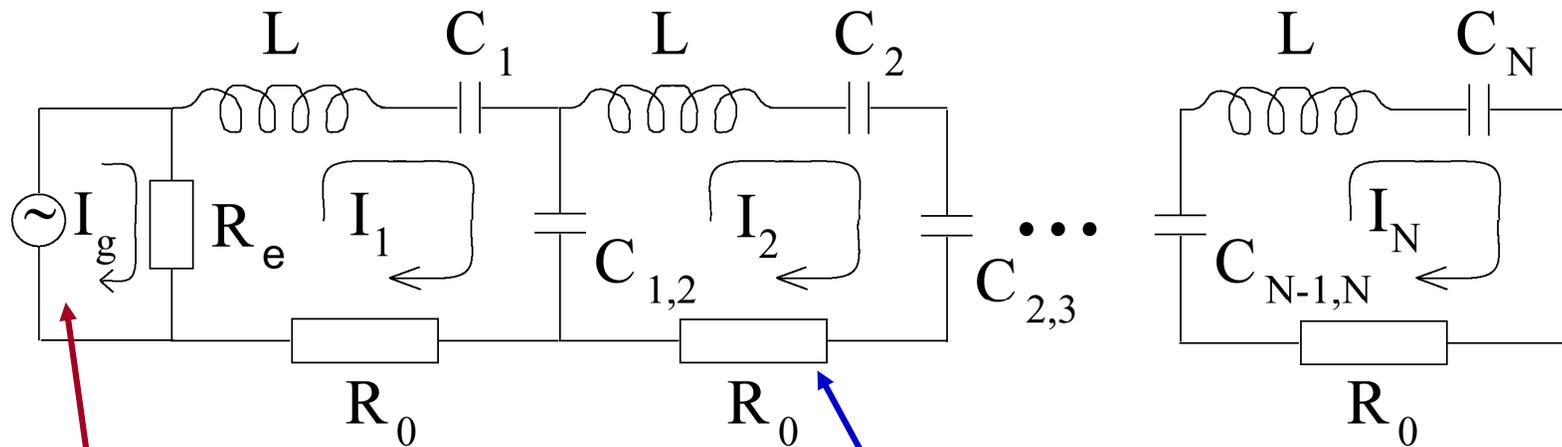
TM010 Eigenmodes Example: 9-Cell Cavity (2)



*accelerating mode: π cell-
to-cell phase advance*



Circuit Model: Step 7 Add Losses and a Generator



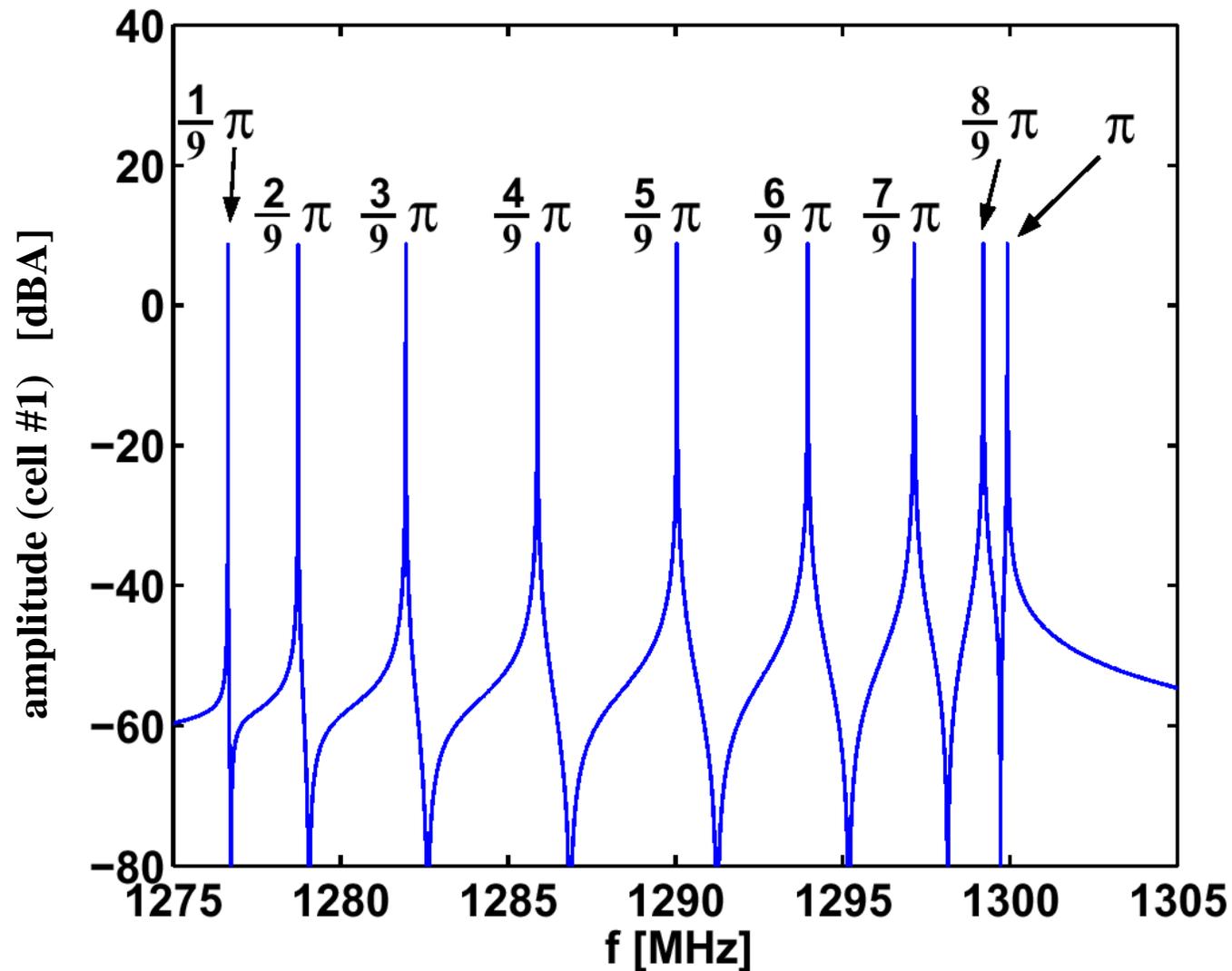
**generator
current**

losses:

$$Q_n = \frac{\omega L}{R_0} = \frac{\omega U_n}{P_{dis,n}}$$

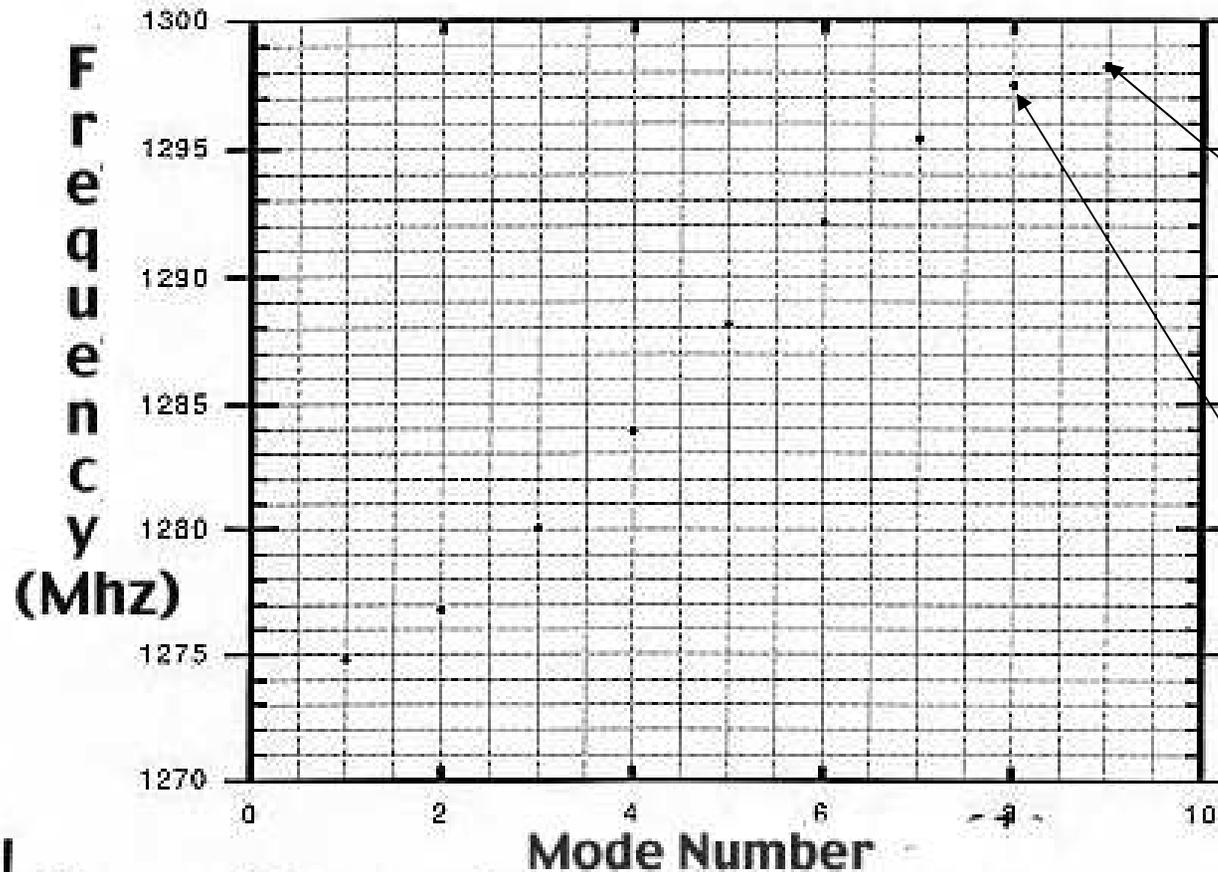


Simulation Example: TM₀₁₀ Eigenmode-Spectrum of a 9-Cell Cavity

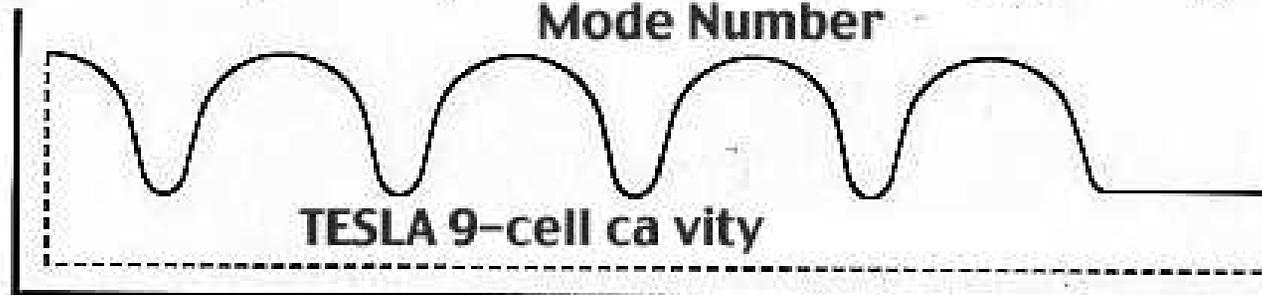




Dispersion Relation



The working point. If it is too close to the neighbor point this neighbor mode can also be excited. To avoid this, more cell-to-cell coupling is needed: broader aperture.





Mode Beating during Cavity Filling

► Modeling of the transient state (mode beating)

Example: 7-cells, $k_{cc}=1.85\%$, $Q_L=3.4 \cdot 10^6$

